

## CHAPTER : 5

### EFFECT OF DAMPING RESISTOR ON CRITICAL FLASHOVER

#### VOLTAGE $U_{50}$ AND VOLTAGE DISTORTION ON AC WAVE

##### 5.1 General

The method of measurement of voltage distribution across the test object is discussed in detail in Chapter Number-4. The transferred impulse voltage on AC terminal of the test object was measured at points A and B using damped capacitive voltage divider as well as resistive voltage divider. As discussed in Chapter-4, in the case of capacitive controlled rod-rod gap, some difficulties were faced in the measurement of lightning impulse transferred voltage  $U_A$ . Hence, it was felt necessary to adopt a method of indirect calculation of voltage  $U_A$ . The magnitude of voltage  $U_A$  should be derived without connecting any divider, because, once an element is connected between terminal A and earth, the circuit configuration and response of the circuit to fast rising pulse changes and hence on account of self and stray capacitance of the divider the measured value of voltage  $U_A$  may not be accurate. Thus it was a must to verify the accuracy of the measured voltage  $U_A$ . It is more important to verify this for the case of capacitive controlled rod-rod gap because large difference exists between voltage  $U_A$  and  $U_B$ . Because of this large difference the actual voltage stress across the longitudinal insulation will be less. In the case of rod-rod gap the difference between voltages  $U_A$  and  $U_B$  is very small. Hence, if the damping resistor is connected between the terminals of the test

object and the supporting capacitance  $C_2$ , then the voltage stress across the open gap is not affected much in the case of rod-rod gap.

The indirect method adopted for estimating voltage  $U_A$  is to derive critical flashover voltage  $U_{50}$ , with and without damping resistor  $R$  for  $\alpha$  equal to zero ( $\alpha = U_{AC}^- / U_I^+ + U_{AC}^-$  where  $U_I^+$  is impulse voltage and  $U_{AC}^-$  is peak value of AC voltage, if  $U_{AC}^-$  is equal to zero,  $\alpha$  is zero). To derive  $U_{50}$  of a given test object it is not necessary to connect any divider at point A and hence the stray capacitance to earth does not play any role in deriving  $U_{50}$  voltages. The most popular Up and Down method was used to derive  $U_{50}$  voltage of the test object. This method is described in this section.

The measurement of critical flashover voltage and standard deviation sigma was done for capacitive controlled rod-rod gap and rod-rod gap electrode configuration. The results are analysed and discussed at depth in the following sections. Comparison of measured and calculated transferred impulse voltages is also discussed.

If the damping resistor is connected between the terminals of the test object and the supporting capacitance  $C_2$ , it introduces an error in the measurement of transferred impulse voltage (voltage  $U_B$ ) on AC terminal and hence this leads to an error in the compensation of the voltage drop on AC terminal. Also the voltage distribution across the test object is affected. These two phenomena leads to an

error in deriving actual  $U_{50}$  voltage of the test object. The method of calculation of total reduction of voltage stress across the test object because of such type of circuit connections is discussed.

Finally, if it is a must to connect a damping resistor between the terminals of the supporting capacitor and the High Voltage testing transformer, the question that arises is, how to protect the high voltage testing transformer? The time constant of the circuit can be increased either by increasing the value of the damping resistor or by connecting an additional capacitor across the terminals of the testing transformer. The effect of additional capacitor ( $C_3$ ) on the voltage distribution across the test object was also studied and the results are discussed in this chapter.

## 5.2 Up and Down method/149/

The Up and Down method is more and more frequently used in dielectric testing because of its simplicity and accuracy for  $U_{50}$ . In IEC publication 60, only the estimate of  $U_{50}$  is discussed. But there also exist estimates of sigma, and of the variances of  $U_{50}$  and sigma/151, 152, 153/.

It is widely known that in comparison with the interpolation method, the Up and Down method enables to estimate  $U_{50}$  with smaller number of voltage applications when the estimation of reliability is assumed to be the same/149/.

Various laboratories have used different methods like in-

interpolation method, Up and Down method, extended Up and Down method and sequential Up and Down method. Digital computer simulation technique was used to clarify the relation between testing method and estimation of reliability. It was clearly shown both theoretically and experimentally that when the number of test voltage application is finite, the estimate of  $\sigma$  obtained by the interpolation method is greater than the true value and that obtained by the Up and Down method is smaller/149,152/.

Having chosen a voltage step called  $d$  (also known as  $\Delta U$ ) this method consists of applying a series of test voltages. The result of application of any impulse determines the voltage for the next one; if it is a breakdown, the next impulse voltage is applied at one step lower, and if it is a withstand, one step higher. Figure 5.1 explains the modality of this method. The first significant voltage level is selected as being the first one giving a result different from the preceding series. Only this one and the following ones are taken under consideration in the calculations. These so called "significant tests" should be made at voltage  $U_k$ ,  $k$  being the index of the voltage levels, the first one corresponding to  $k=0$  is  $U_0$ . The estimation proposed by Dixon and Mood/150/consists in forming  $n_k^0$  and  $n_k^1$  (the total number of withstands and breakdowns at each voltage level  $U_k$ ), calculating,

$$N_0 = \sum_k n_k^0 \quad \text{and}$$

$$N_1 = \sum_k n_k^1 \quad \text{and then working with only the}$$

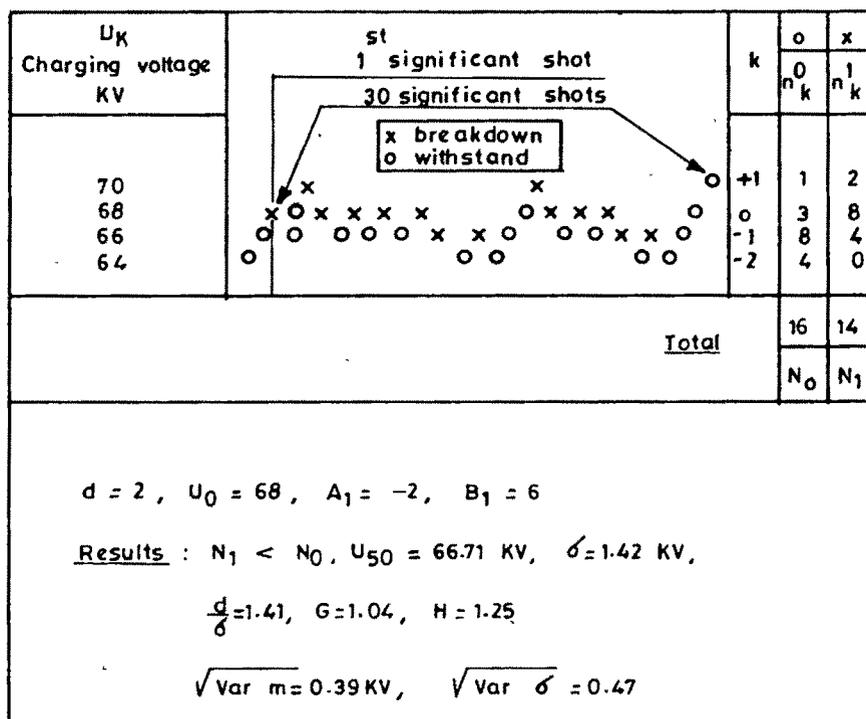


Figure. 5-1 Typical example of 30 shots applied according to the modalities of the "Up and Down" method (ref.149 ).

$n_k^0$  or  $n_k^1$  values giving the smaller of  $N_0$  and  $N_1$ .

The estimates of  $U_{50}$ , sigma and their standard errors are given by the formulas: ("^" read as "estimate of"):

If  $N_1 \leq N_0$ ,

$$\hat{U}_{50} = U_{0+d} \times \left[ \frac{A_1}{N_1} - \frac{1}{2} \right] \dots\dots\dots (5.1)$$

$$\hat{\sigma} = 1.620 \times d \left[ \frac{N_1 \times B_1 - A_1^2}{N_1^2} + 0.029 \right] \dots\dots\dots (5.2)$$

with

$$A_1 = \sum_k k n_k^1$$

$$B_1 = \sum_k k^2 n_k^1$$

$$\sqrt{\text{var}(U_{50})} = G \times \hat{\sigma} / \sqrt{N_1} \dots\dots\dots (5.3)$$

$$\sqrt{\text{var}(\hat{\sigma})} = H \times \hat{\sigma} / \sqrt{N_1} \dots\dots\dots (5.4)$$

If  $N_0 < N_1$ ,

$$\hat{U}_{50} = U_{0+d} \times \left[ \frac{A_0}{N_0} + \frac{1}{2} \right] \dots\dots\dots (5.5)$$

$$\hat{\sigma} = 1.620 \times d \left[ \frac{N_0 \times B_0 - A_0^2}{N_0^2} + 0.029 \right] \dots\dots\dots (5.6)$$

with  $A_0 = \sum_k k n_k^0$

$B_0 = \sum_k k^2 n_k^0$

$$\sqrt{\text{var}(U_{50})} = G \times \hat{\sigma} / \sqrt{N_0} \dots\dots\dots (5.7)$$

$$\sqrt{\text{var}(\hat{\sigma})} = H \times \hat{\sigma} / \sqrt{N_0} \dots\dots\dots (5.8)$$

G and H can be derived from Figure 5.2

In practice, with some tens of significant impulses, the

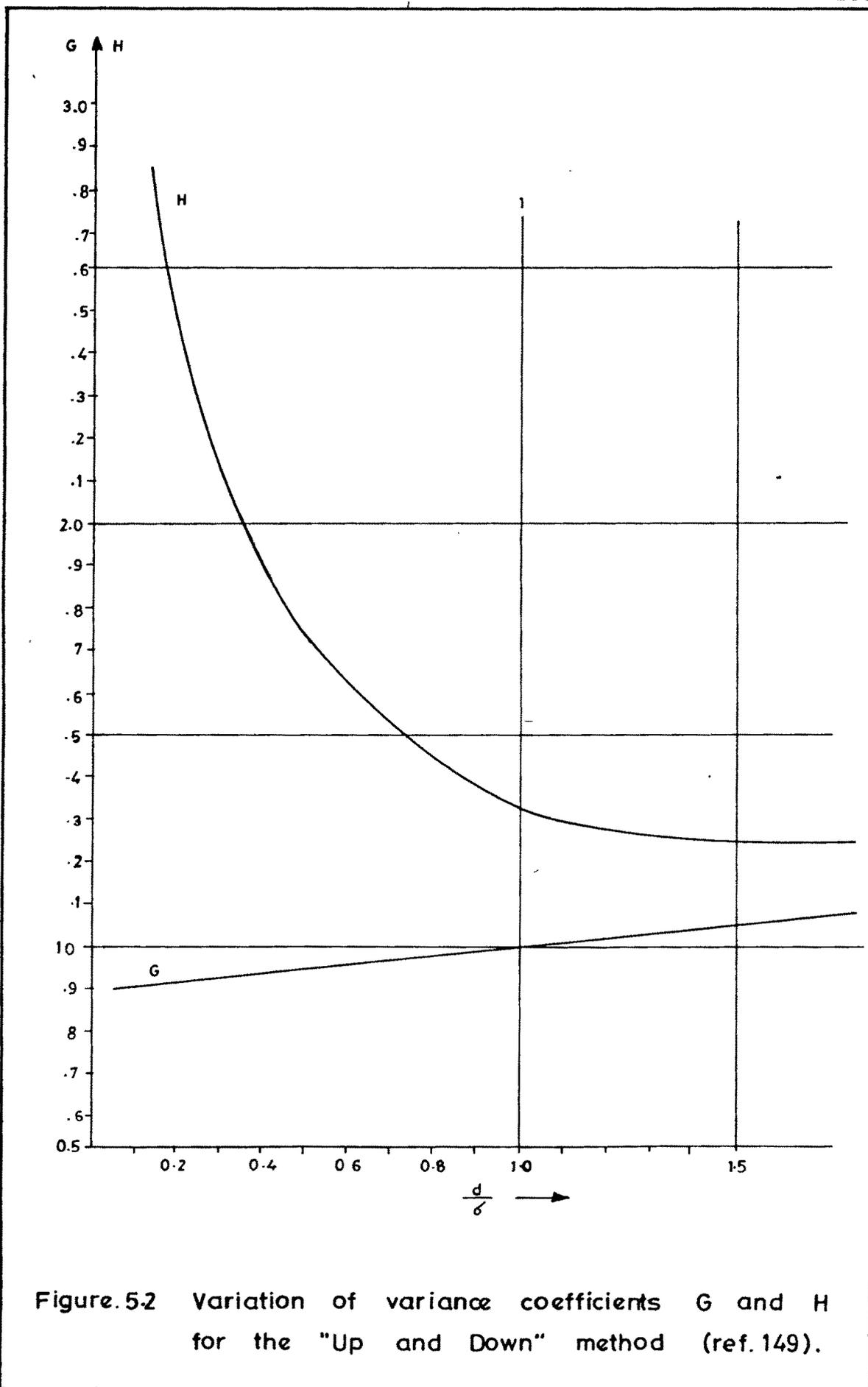


Figure.5.2 Variation of variance coefficients G and H for the "Up and Down" method (ref.149).

formulas are only correct if

$$\frac{N_1 \cdot B_1 - A_1^2}{N_1^2} \quad \text{OR} \quad \frac{N_0 \cdot B_0 - A_0^2}{N_0^2} > 0.3$$

$$\text{And } 0.5 < \frac{d}{\sigma} < 1.7 \quad \dots\dots\dots(5.9)$$

As it can be seen from figure 5.2 it is desirable that the voltage step 'd' should be roughly equal to  $\frac{\lambda}{\sigma}$ .

These limits are respected if after ten or more significant impulses, the number of  $U_k$  values used is 3,4 or 5. Otherwise  $d/\sigma$  is probably out of the limits specified and must be modified/149/.

### 5.3 Measurement of critical flashover voltage and sigma

As discussed in the earlier section the aim of this study was to estimate the transferred impulse voltage  $U_A$  accurately, and to study the effect of damping resistor and supporting capacitance  $C_2$  on the critical flashover voltage  $U_{50}$  of the rod-rod gap and capacitive controlled rod-rod gap. The circuit connection shown in figure 4.1 were used for the measurement of voltage  $U_{50}$ . The applied impulse voltage to terminal D of Figure 4.1 was measured using damped capacitive voltage divider (2). The gap spacing used for this study was not of much importance as the aim was only to study the effect of various elements which are required while conducting bias voltage test for the protection of the high voltage testing transformer from transient overvoltages on  $U_{50}$  and standard deviation sigma.

It has been reported in the literature/151,152/ that in the case of Up and Down method 95% confidence limit of  $U_{50}$  can be easily obtained when number of applied impulses are more than or equal to 30. Hence, to get required accuracy of the results for each combination of damping resistance and supporting capacitance fifty impulses at 25 seconds intervals were applied to the test object to derive  $U_{50}$  values. The voltage step  $d$  (also known as  $\Delta U$ ) for lightning impulses was taken from 2 to 3% of the approximately estimated value of  $U_{50}$  and for switching impulses the voltage step was taken from 4 to 6% of the approximately estimated value of  $U_{50}$ .

The laboratory conditions and layout of the various equipments shown in Figure 4.1 were maintained more or less same to obtain a set of sufficient results. The atmospheric parameters were recorded at 10 minutes interval. Average value of the atmospheric parameters were taken for applying correction factors  $K_d$  and  $K_h$  according to IEC publication-60. The measured 50% discharge voltages  $U_{50}$  reported in the following sections are corrected to those applicable for standard reference atmosphere by dividing by  $K_d/K_h$ . The standard reference atmosphere is:

Temperature	$t_o$	=	20°C
Pressure	$b_o$	=	1013 mbar
Humidity	$H_o$	=	11g water per $m^3$ .

### 5.3.1 Capacitive controlled rod-rod gap

The layout of the capacitive controlled rod-rod gap is

shown in Figure 4.9a. The Up and Down method was used to derive  $U_{50}$  and sigma, which is already discussed in detail in section 5.2.

#### 5.3.1.1 Lightning impulse voltage

The complete bias test circuit was arranged according to Figure 4.1. Lightning impulses having a waveshape of 1.38/56 microseconds were applied to the impulse terminal D of Figure 4.1. The power frequency voltage was not applied, i.e.  $\omega$  is equal to 0.0. Fifty impulses were applied to the test object at 25 seconds interval to derive  $U_{50}$ .

Figure 5.3 gives voltage  $U_{50}$  as a function of damping resistance  $R$  for various values of supporting capacitance  $C_2$ , the test object capacitance  $C_1$  was 600pF. Figure 5.4 gives the same results when test object capacitance  $C_1$  is 1200pF (only one electrode gap shown in Figure 4.9a was used for  $C_1 = 1200\text{pF}$ , the open gap spacing of rod-rod gap was 550mm). For both the cases  $U_{50}$  increases with the increase in the value of damping resistance and it decreases with the increase in the value of supporting capacitance  $C_2$ . To highlight this phenomena more effectively, percentage increase in  $U_{50}$  with respect to damping resistance is calculated, the results are given in Figure 5.5 for the test object capacitance equal to 600pF. The percentage increase of  $U_{50}$  in case of lightning impulse (LI) voltage is as high as 23% when a damping resistance of 8.56Kilo Ohms is connected between the terminals of the test object and the supporting capacitance (for  $C_2$  equal to 19800pF). Since the curves are nonlinear, it is difficult to say what happ-

## Capacitive controlled rod - rod gap (figure.4.9 a)

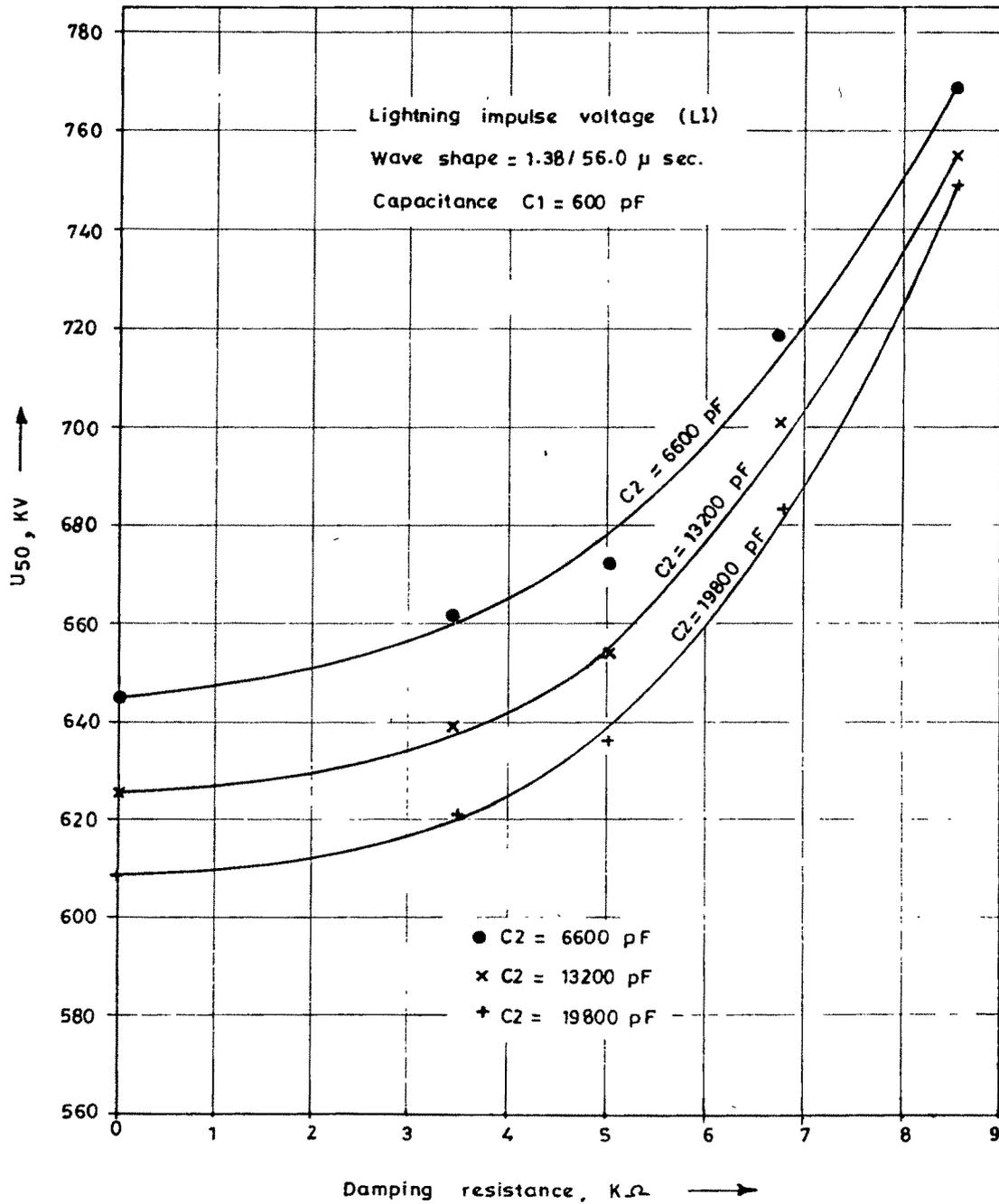


Figure. 5.3  $U_{50}$  (LI) as a function of damping resistance,  
 $C_1 = 600$  pF.

## Capacitive controlled rod rod gap (figure 4.9 a)

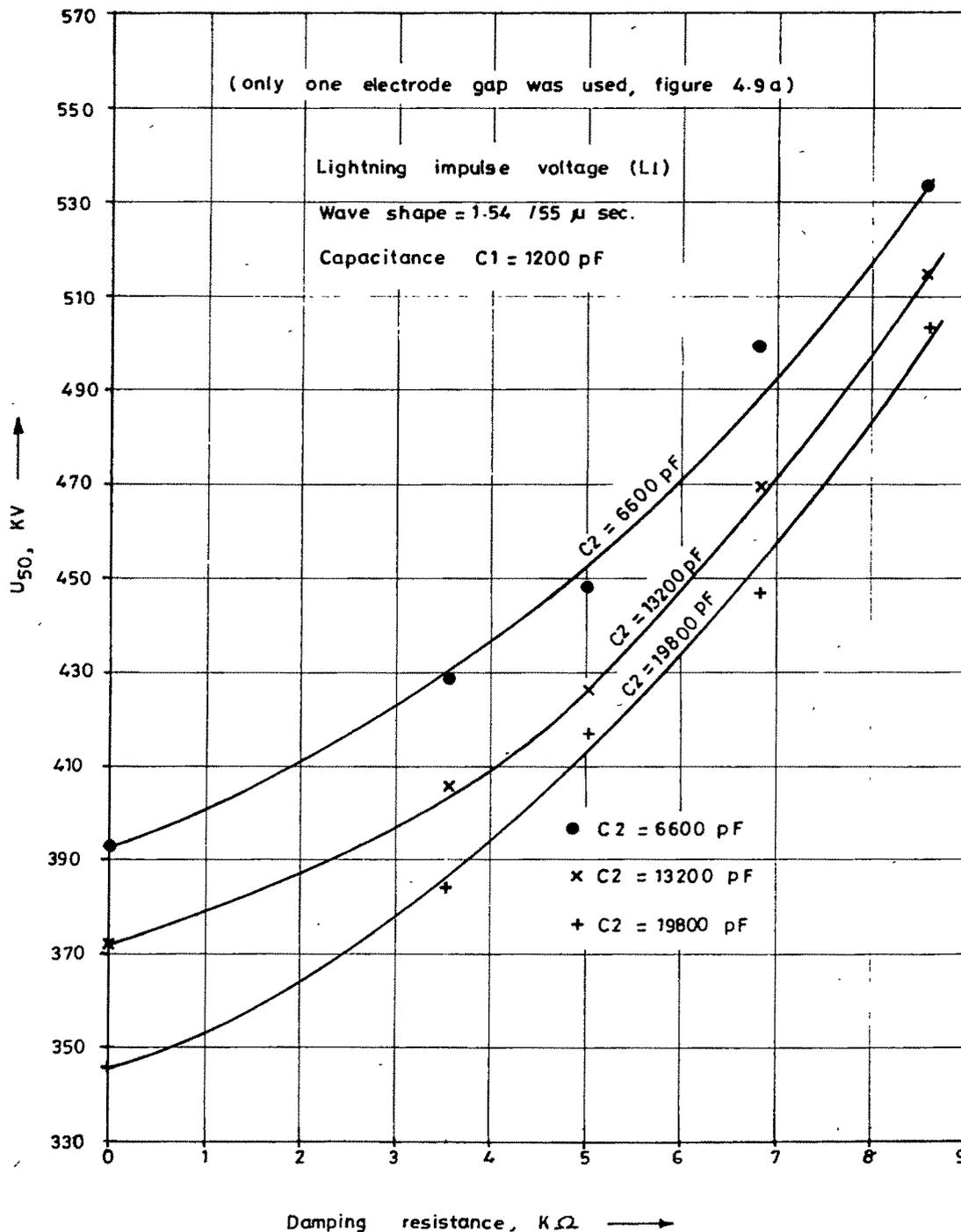
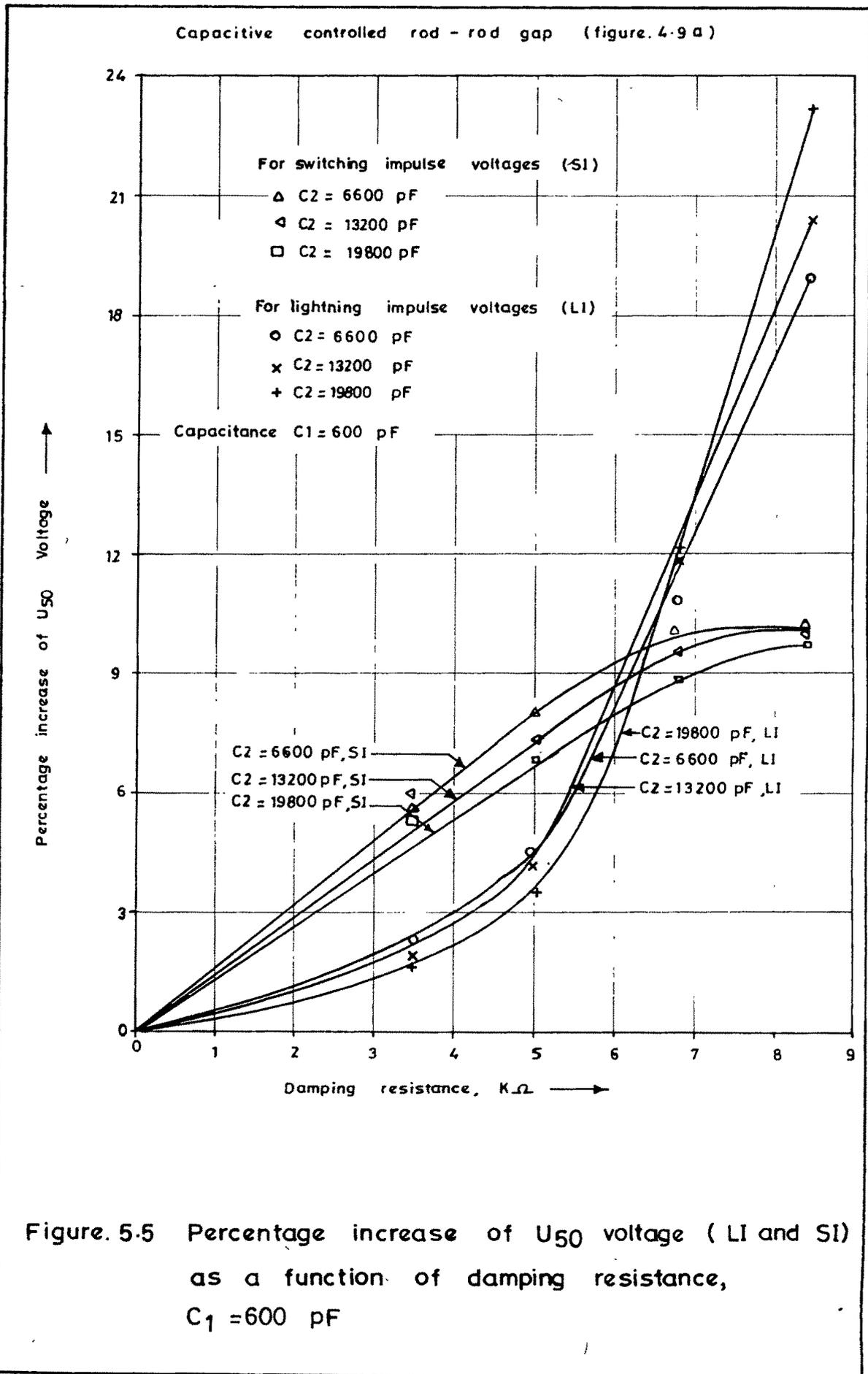


Figure. 5.4  $U_{50}$  (LI) as a function of damping resistance,  
 $C_1 = 1200$  pF. .



ens when the value of damping resistance  $R$  is large. Of course one can get an idea about the error introduced in deriving the actual critical flashover voltage or withstand voltage of the test object by not connecting the supporting capacitance  $C_2$  directly at the terminal of the test object.

#### 5.3.1.2 Switching impulse voltage

The same capacitive controlled rod-rod gap shown in Figure 4.9a was used to derive  $U_{50}$  for various values of damping resistance and supporting capacitance  $C_2$ . Switching impulse wave shape was 250/2750 microseconds. Fifty impulses were applied for each case at 25 seconds interval. The atmospheric parameters were recorded at 10 minutes intervals.

Figure 5.6 gives  $U_{50}$  as a function of damping resistance for various values of supporting capacitance  $C_2$  for test object capacitance equal to 600pF. The increase in  $U_{50}$  is nonlinear with the increase in damping resistance  $R$ . The curves obtained for three different values of supporting capacitance  $C_2$  are parabolic and parallel to each other. Figure 5.7 gives the same results for test object capacitance equal to 1200pF. The switching impulse voltage wave shape for  $C_1=1200\text{pF}$  was 259/2700 microseconds. The percentage increase of  $U_{50}$  with increase in damping resistance is shown in Figure 5.5 for  $C_1=600\text{pF}$ . The percentage increase in  $U_{50}$  for three values of supporting capacitances 6600pF, 13,200pF and 19,800pF is nearly same for a given value of damping resistance. The increase in  $U_{50}$  is appro-

Capacitive controlled rod - rod gap (figure. 4.9 a)

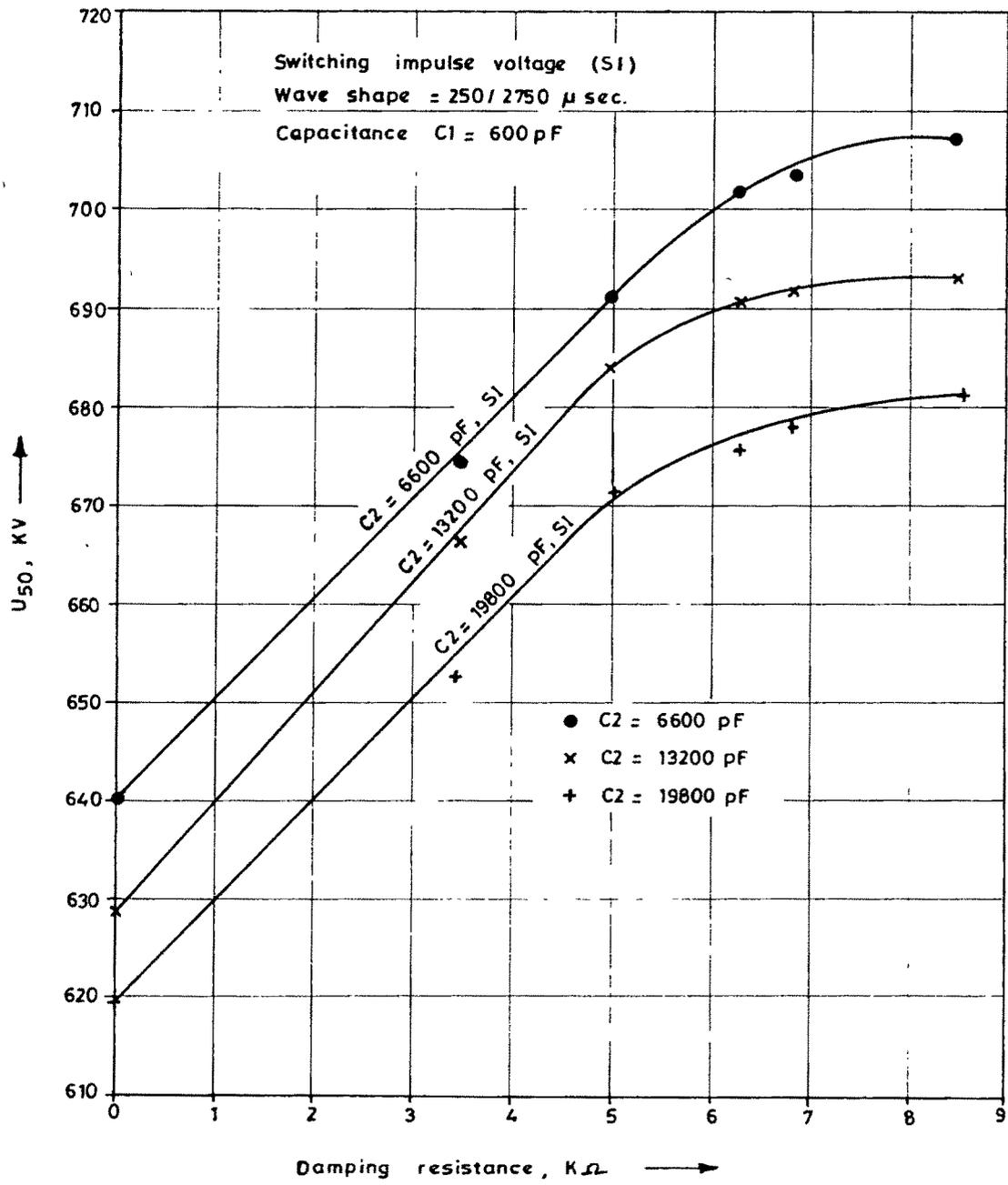


Figure.5.6  $U_{50}(SI)$  as a function of damping resistance,  
 $C_1 = 600$  pF.

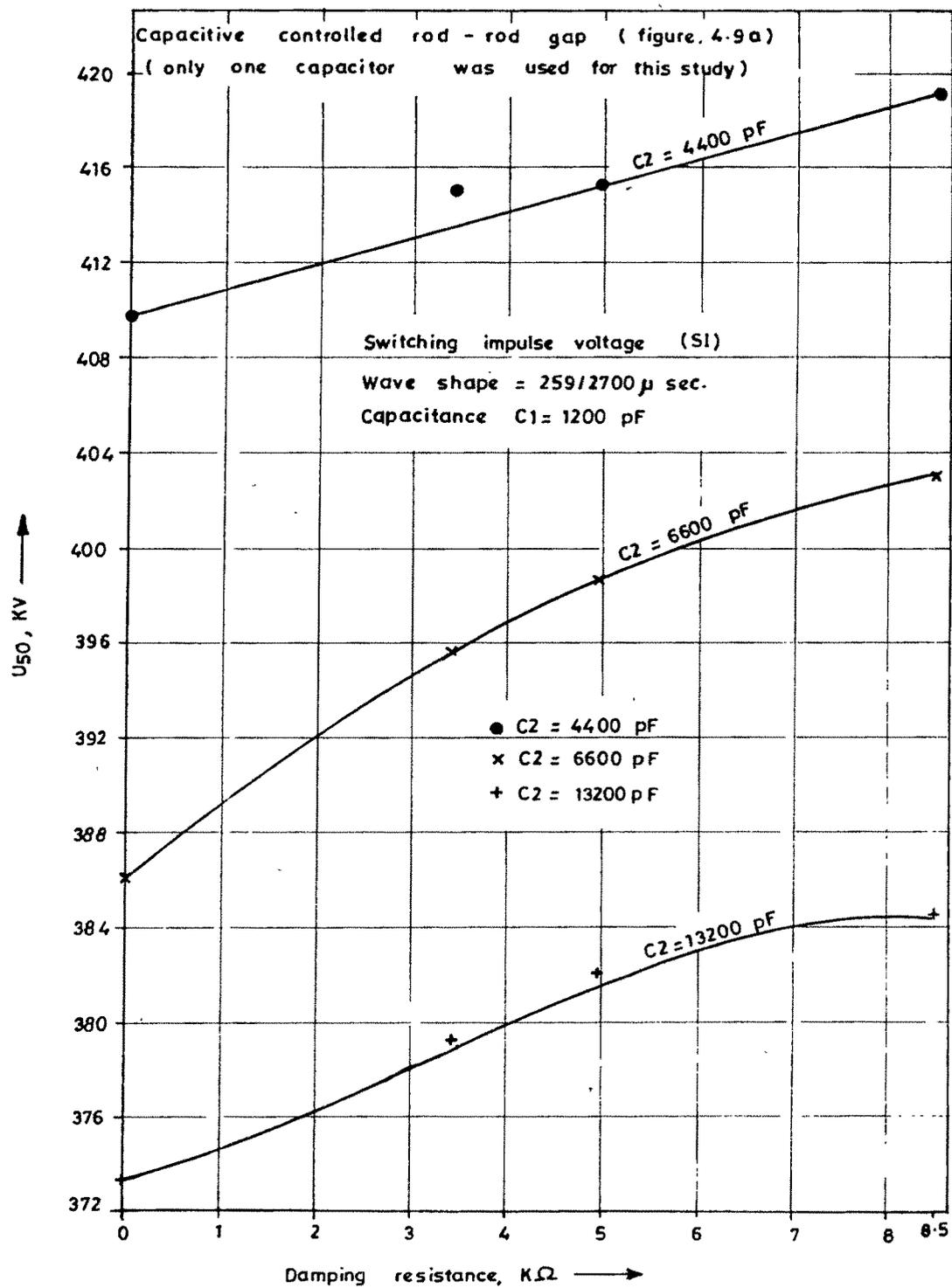


Figure. 5.7 U<sub>50</sub>(SI) as a function of damping resistance,  
 $C_1 = 1200$  pF.

approximately 10% when damping resistance of 8.56 Kilo Ohms is connected between the terminals of the test object and the supporting capacitance  $C_2$ . This is approximately 2.3 times less compared to the case when lightning impulse voltage is applied. As seen from Figure 5.5, the percentage increase in  $U_{50}$  for damping resistance greater than 7 Kilo Ohms is very small. Almost a saturation has reached for damping resistance  $R$  equal to 7.0 Kilo Ohms, i.e. further increase in damping resistance does not lead to increase in  $U_{50}$ . However, it might be necessary to verify the characteristics of the curves for damping resistance greater than 8.56 Kilo Ohms. Since maximum available value of high voltage damping resistance was only 8.56 Kilo Ohms, further tests could not be conducted.

### 5.3.2 Rod-rod gap

#### 5.3.2.1 Lightning impulse voltage

The electrode arrangement shown in Figure 4.9b was used to derive  $U_{50}$  of rod-rod gap having an open gap spacing of 770mm. The wave shape of the lightning impulse voltage was 1.08/47.0 microseconds. Fifty impulses were applied to the electrode system at 25 seconds interval to obtain  $U_{50}$  for each test set up conditions. The voltage step was equal to 3% of the estimated  $U_{50}$  voltage. Figure 5.8 gives  $U_{50}$  voltage as a function of damping resistance for various values of supporting capacitance. The curves are more or less linear, i.e. the value of  $U_{50}$  increases linearly with the increase in damping resistance. Many times it was also

Rod - Rod gap Configuration (figure. 4-9 b)

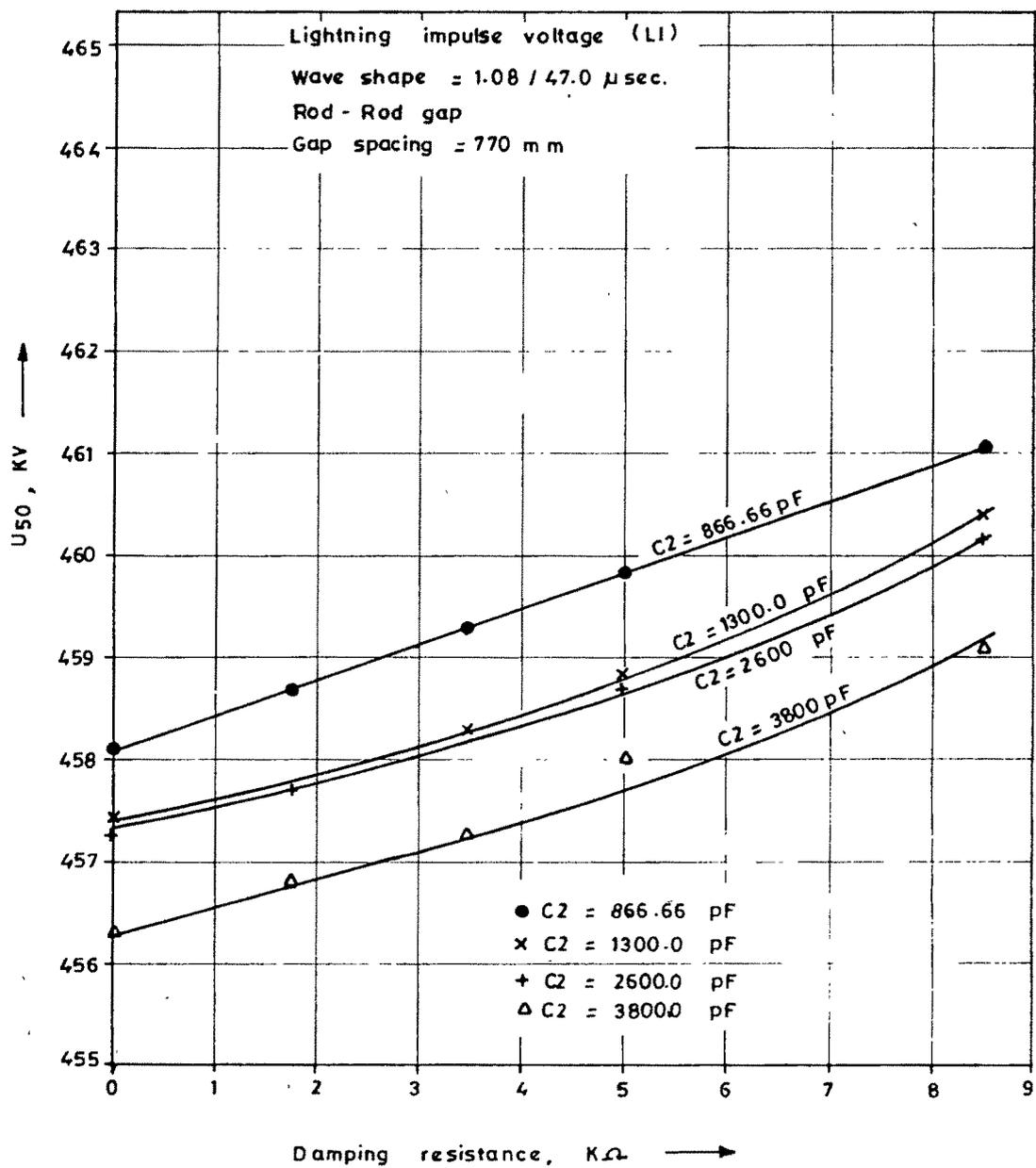


Figure. 5-8 U<sub>50</sub> (LI) as a function of damping resistance, rod-rod gap.

Observed that  $U_{50}$  for higher value of damping resistance is lower than that obtained for lower value of damping resistance (the difference was within  $\pm 1.5$  to  $\pm 3.0$  kV), theoretically this is not correct and hence many times the test had to be repeated even three to four times to get the correct value of  $U_{50}$  i.e. the value of  $U_{50}$  should be higher for higher value of damping resistance or it should be at least equal but not less. This difference may be because of the statistical variation of  $U_{50}$ . It is worthwhile mentioning here that, when the test was conducted for various values of supporting capacitance  $C_2$  and damping resistance  $R$ , the atmospheric parameters were not same because of change in day and time. Hence, this phenomena may be also attributed to the variations in the correction factors  $K_d$  and  $K_h$  given by IEC publication-60.

Figure 5.9 gives the percentage increase of  $U_{50}$  as a function of damping resistance. The percentage increase in  $U_{50}$  is maximum 0.7 per cent when damping resistance of 8.56 kilo ohms is connected between the terminals of the test object and the supporting capacitance. This is negligible compared to the case of capacitive controlled rod-rod gap. Hence, it may be concluded that in the case of disconnector type open gaps, (which represents rod-rod gap) the placement of a damping resistance in the bias test circuit is not important. It may be connected between the terminals of the disconnector and the supporting capacitance  $C_2$ , so that the severity of the chopped wave is not dangerous for the winding insulation of High Voltage power frequency

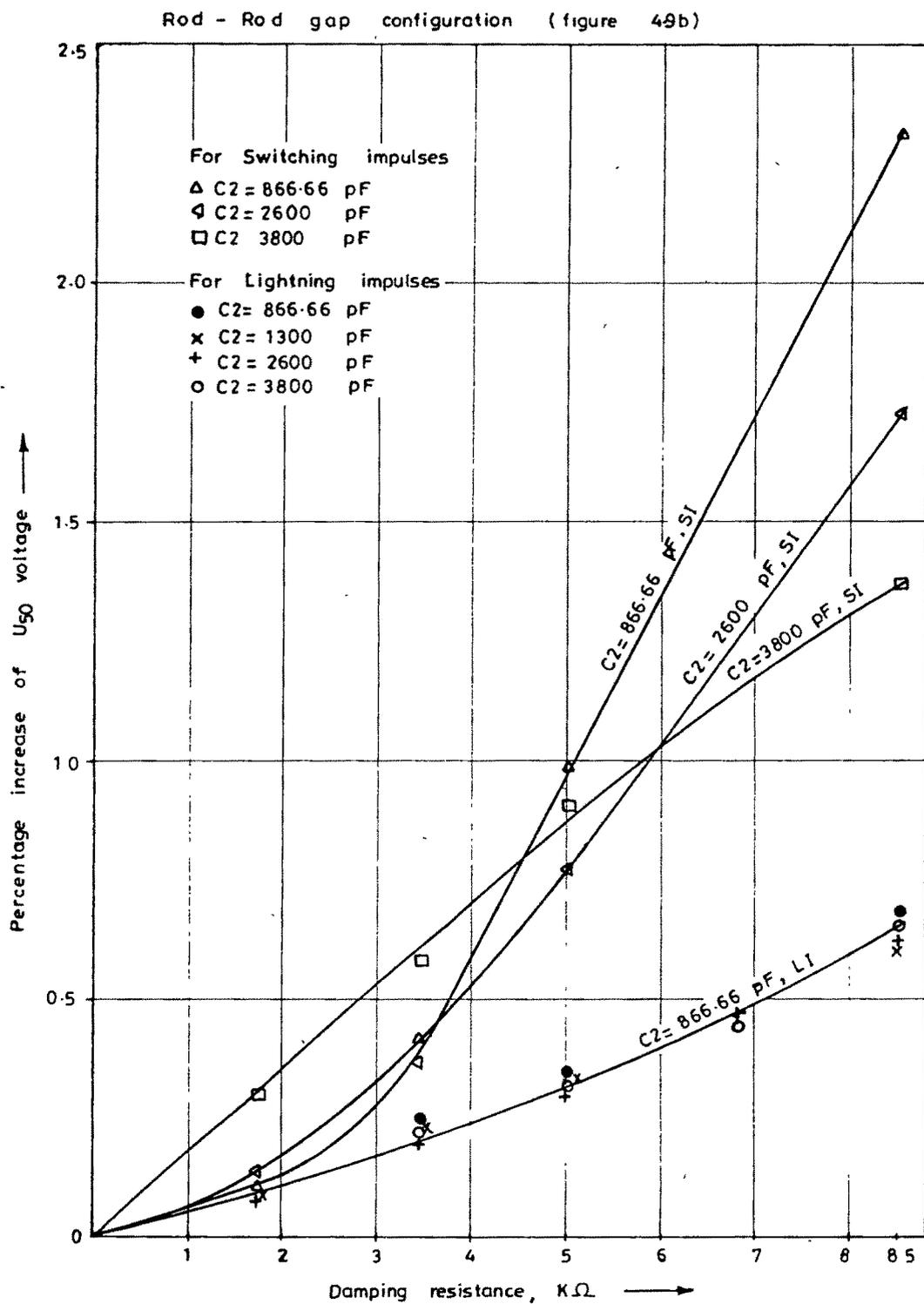


Figure 5.9 Percentage increase of  $U_{50}$  voltage (LI and SI) for rod-rod gap as a function of damping resistance.

test transformer.

### 5.3.2.2 Switching impulse voltage

The same electrode configuration, shown in Figure 4.9b, was used to derive  $U_{50}$  of rod-rod gap but having an open gap spacing of 500mm. The open gap had to be reduced from 770mm to 500mm because of flashovers occurring from live electrode to earth while deriving 50% discharge voltage  $U_{50}$ . It was decided to reduce the open gap clearance instead of increasing the height of the horizontal rod-rod gap electrode configuration to prevent flashovers to earth. The switching impulse voltage waveshape was 238/2655 microseconds. Fifty impulses were applied at 25 seconds interval to the impulse terminal D of Figure 4.1 at voltage step  $\Delta U$  equal to sigma (i.e. 6% in case of switching impulses). Figure 5.10 gives  $U_{50}$  voltage as a function of damping resistance for various values of supporting capacitance. The difficulty which was encountered in the measurement of  $U_{50}$  voltage in case of lightning impulse voltage was also noticed in the case of switching impulses and hence the test had to be repeated for certain values of damping resistance and capacitance  $C_2$ . The percentage increase of  $U_{50}$  voltage is given in Figure 5.9. The percentage increase in  $U_{50}$  is between 1.3 and 2.3 per cent when damping resistance of 8.56 kilo Ohms is connected between the test object terminal and the supporting capacitance  $C_2$  ranging from 866.66pF to 3800pF. The conclusion drawn in case of lightning impulse voltage is also valid for switching impulses, and hence in this case also one can connect the damping resistor between the test object ter-

Rod - Rod gap Configuration (figure. 4.9 b )

Switching Impulse Voltage (SI)

Wave shape = 238/2655  $\mu$ sec.

Rod - Rod gap

Gap spacing = 500 mm

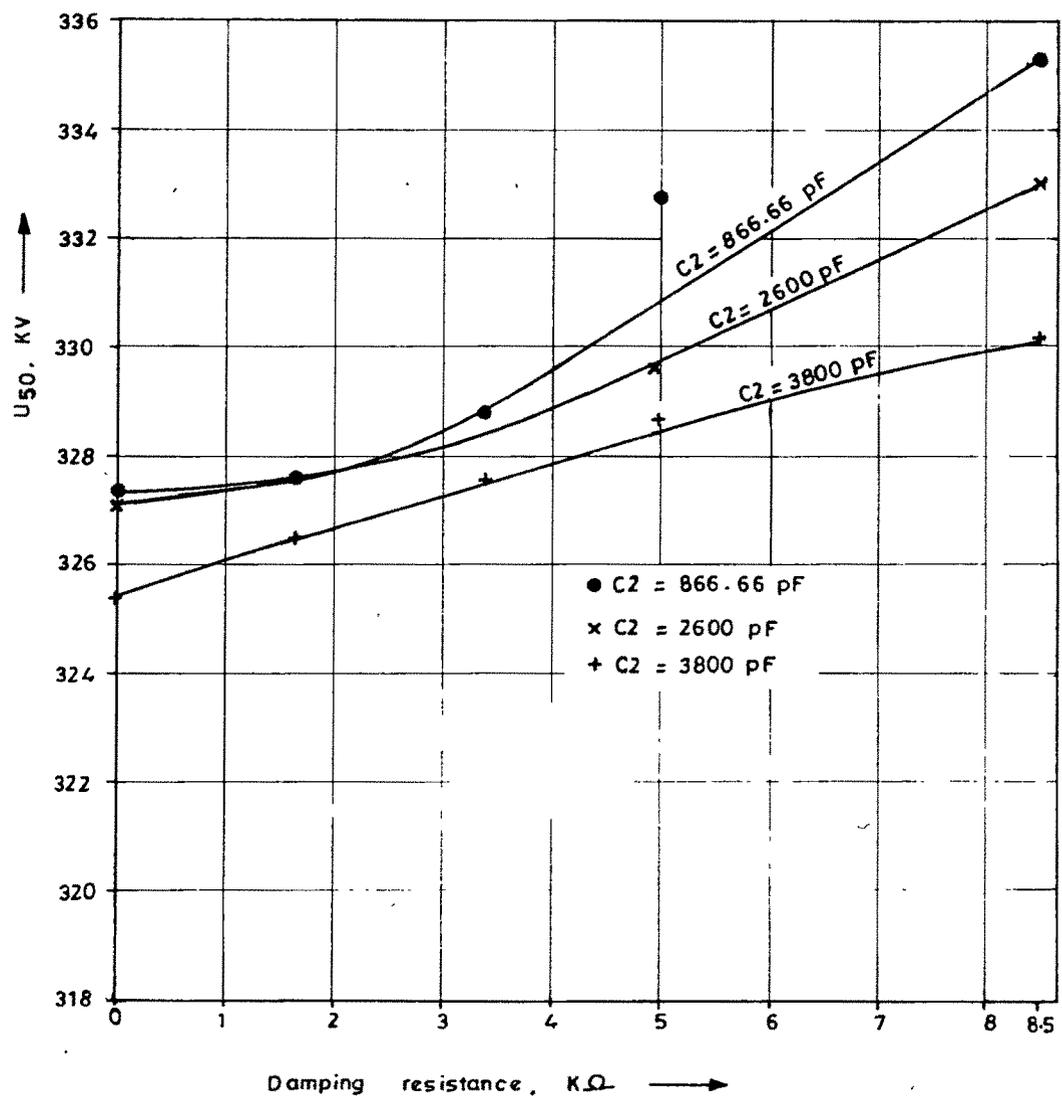


Figure. 5.10  $U_{50}$  (SI) as a function of damping resistance, rod - rod gap.

terminal and the supporting capacitance  $C_2$ .

#### 5.4 Comparison of measured and calculated $U_{50}$

For both the electrode systems, before obtaining  $U_{50}$  with supporting capacitor  $C_2$  and/or damping resistance, first  $U_{50}$  of only rod-rod gap and capacitive controlled rod-rod gap configuration was obtained by earthing point A (refer Figure 4.1). This gives  $U_{50}$  of only the test object gap. In other words, this is the test circuit layout used while conducting the equivalent bias voltage test on the test object, i.e. a sum of two voltage components specified by the relevant standards is applied to one terminal the opposite being earthed.  $U_{50}$  voltage was obtained this way and then this value was compared with the  $U_{50}$  voltage of the same test object when connected as per circuit given in Figure 4.2, i.e. Bias Test circuit.

##### 5.4.1 Capacitive controlled rod-rod gap

The  $U_{50}$  for lightning and switching impulse voltages for capacitive controlled rod-rod gap was obtained by earthing point A (refer Figure 4.1), and also when supporting capacitance  $C_2$  is connected in the circuit ( $U_{50}^*$ ). When supporting capacitor  $C_2$  is connected in the circuit, the part of the voltage  $U_{50}^*$  appears across the power frequency terminal of the test object and earth. This voltage,  $U_B$ , can be calculated using the following formula (damping resistance  $R = 0.0$  Ohms, in this case voltage  $U_A$  is same as voltage  $U_B$ ):

$$U_B = \frac{C_1}{C_1+C_2} \times U_{50}^* \quad \dots\dots\dots (5.10)$$

where  $C_1$  = test object capacitance in pF,  $C_2$  = supporting capacitance in pF and  $U_{50}^*$  is the 50% discharge voltage of the test object when supporting capacitor  $C_2$  is connected. Now  $U_{50}$  of only test object (i.e. when terminal 'A' is earthed and  $\alpha = 0.0$ ) is known, and it should remain more or less same. When supporting capacitor  $C_2$  is connected (without any damping resistor, i.e. in this case the damping resistor may be connected between the supporting capacitance and the terminal of the high voltage testing transformer), the percentage of voltage which will appear at the terminal A or B can be calculated. Based on this  $U_{50}^*$  can be derived theoretically as follows:

$$U_{50}^* = \frac{U_{50} \text{ of only gap}}{1 - \frac{C_1}{C_1+C_2}} \quad \dots\dots\dots (5.11)$$

$\frac{C_1}{C_1+C_2}$  gives the per unit voltage  $U_B$  when an impulse voltage is applied to the impulse terminal D of Figure 4.1. By using this philosophy  $U_{50}^*$  voltage was derived theoretically and results are compared. Figure 5.11 gives the calculated  $U_{50}^*$  and measured  $U_{50}^*$  voltage for lightning and switching impulse voltages (the test object capacitance  $C_1$  was 600pF). It is seen from this figure that the measured value of  $U_{50}^*$  is approximately 4.5 per cent more compared to the calculated  $U_{50}^*$  for both lightning impulse and switching impulse voltages. Figure 5.12 gives the measured and calculated

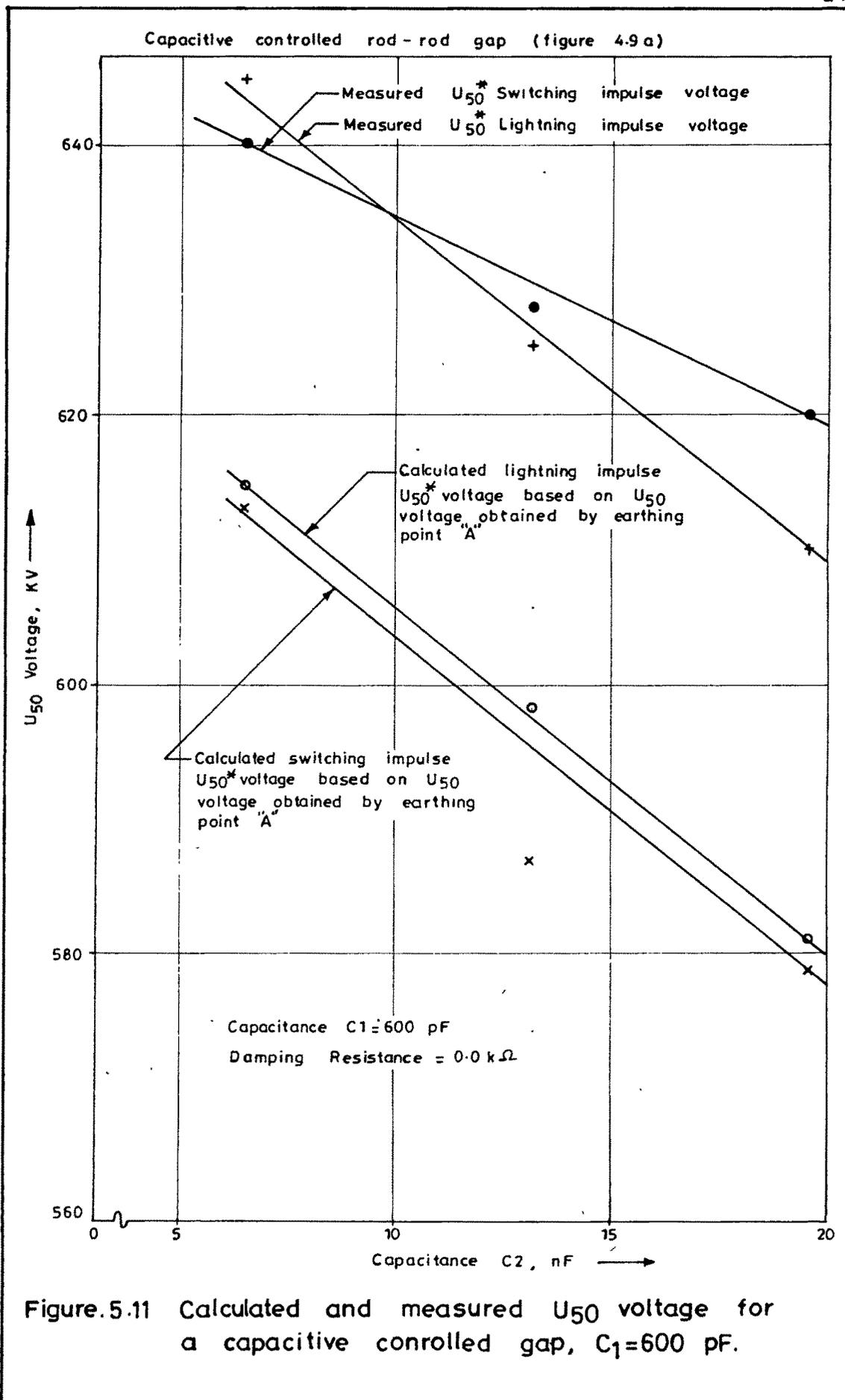


Figure.5.11 Calculated and measured U<sub>50</sub> voltage for a capacitive controlled gap, C<sub>1</sub>=600 pF.

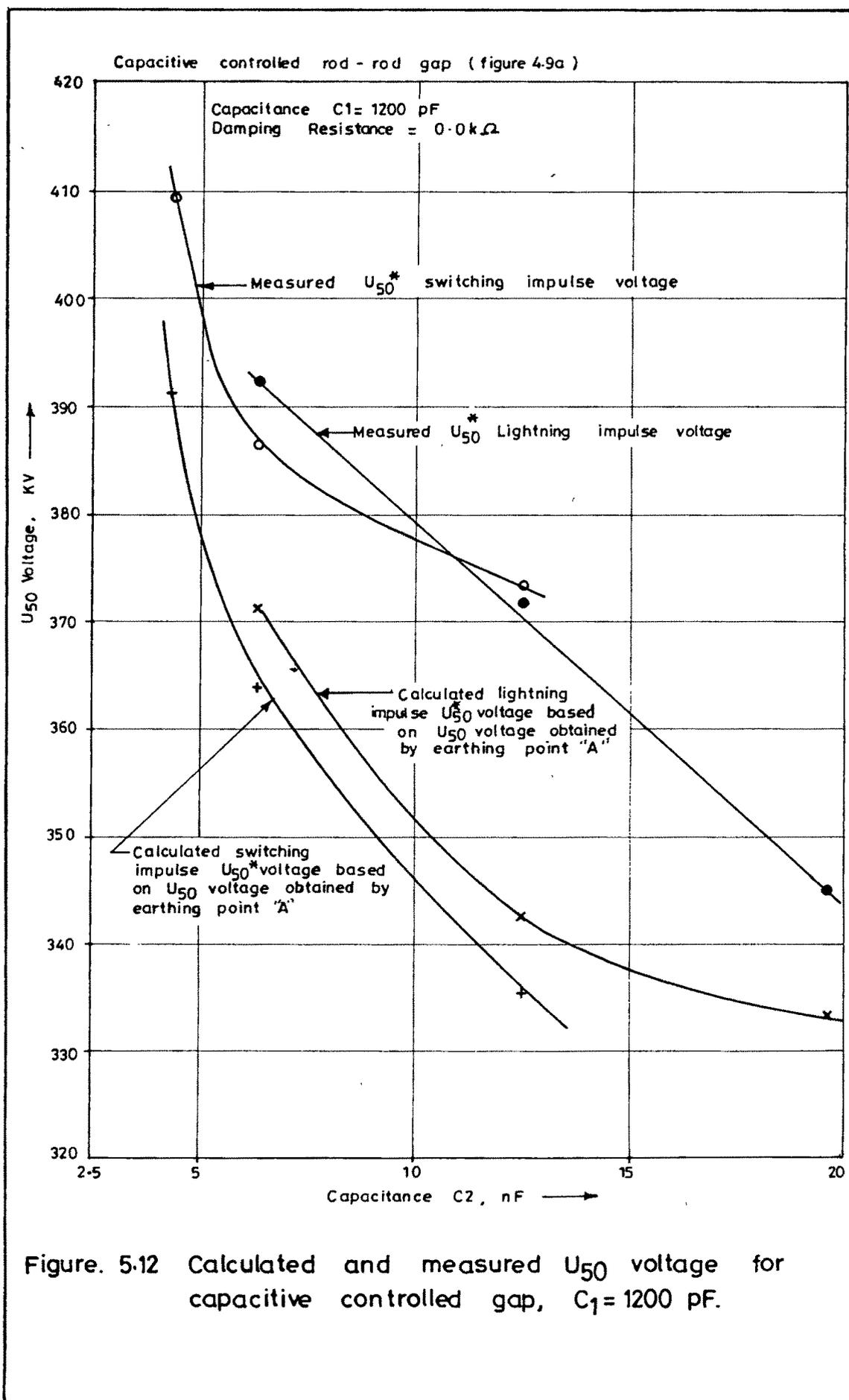


Figure. 5.12 Calculated and measured  $U_{50}$  voltage for capacitive controlled gap,  $C_1 = 1200 \text{ pF}$ .

$U_{50}^*$  voltages when test object capacitance is 1200pF. In this case also the measured values are higher by approximately 4.5 per cent. This difference may be explained as follows: When terminal 'A' is earthed to obtain  $U_{50}$  of only test object gap, then one terminal of the test object is live and other is at earth potential. However, when supporting capacitor  $C_2$  is connected, the other terminal of the test object is earthed through the supporting capacitor  $C_2$ . In this case the other terminal of the test object is not directly at earth potential and hence the electrode field configuration is not same compared to the earlier case. This effect may be also because of the stray capacitance. It may be stated that the measured values of  $U_{50}$  are accurate.

#### 5.4.2 Rod-rod gap

Figure 5.13 gives the measured and calculated values of  $U_{50}$  for various values of supporting capacitance  $C_2$  for rod-rod gap for both lightning and switching impulse voltages. The value of  $U_{50}^*$  voltage for various values of supporting capacitance  $C_2$  was calculated using formula (5.11). The capacitance between two rods was of the order of 11.5pF. For lightning impulse voltages the measured values of  $U_{50}^*$  are approximately 3.8 per cent higher and for switching impulse voltage the measured values of  $U_{50}^*$  are approximately 6.5 per cent higher. The percentage difference between measured and calculated  $U_{50}^*$  remains approximately same irrespective of the value of the supporting capacitance  $C_2$  for both lightning and switching impulse voltages.

Rod - Rod gap Configuration ( figure 4.9 b )

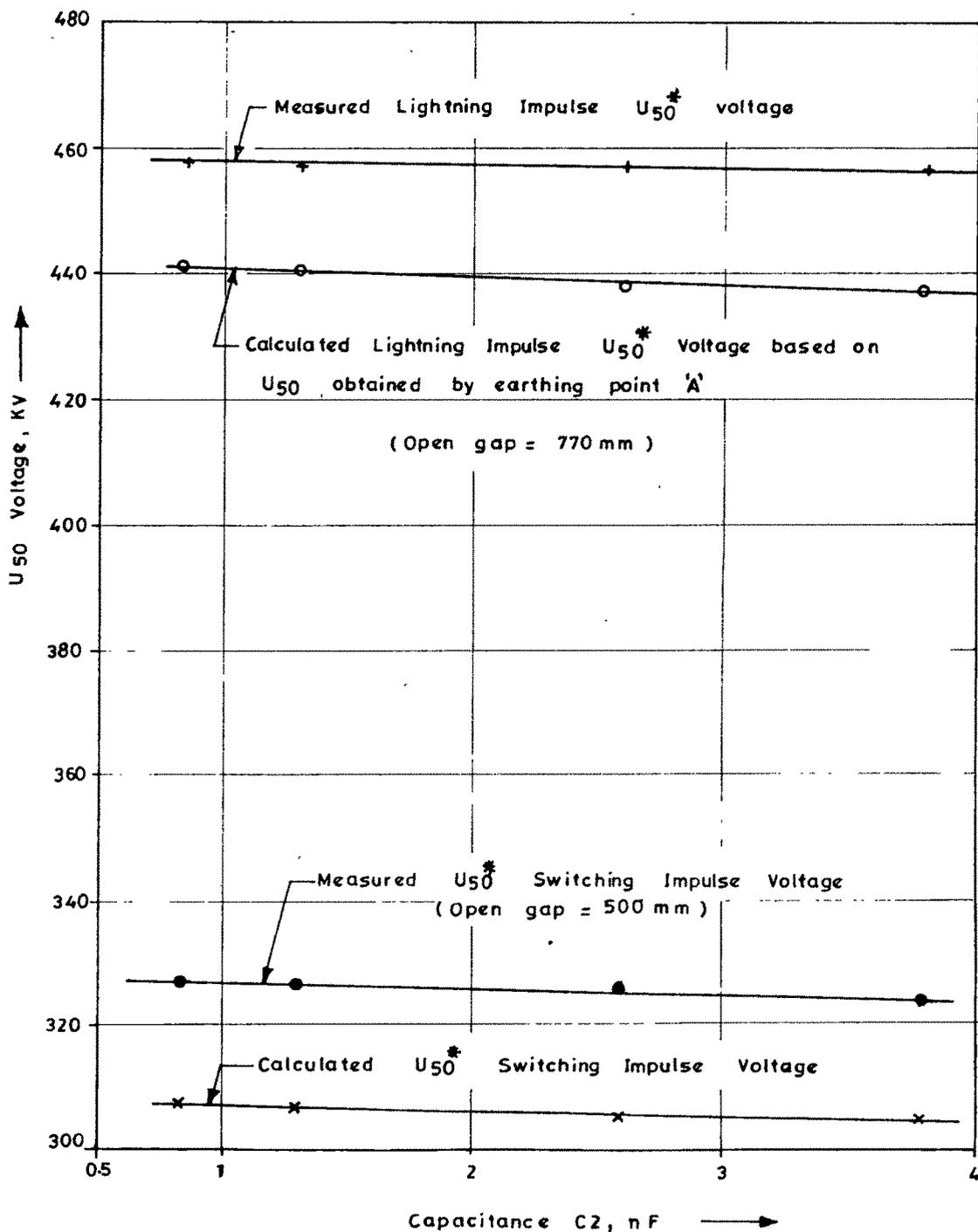


Figure. 5.13 Calculated and measured U<sub>50</sub> Voltage for horizontal Rod - Rod gap.

## 5.5 Comparison of measured and calculated transferred impulse voltage $U_A$

In Chapter-4, the method of measurement of transferred impulse voltages  $U_A$  and  $U_B$  have been discussed in detail. To verify the accuracy of the measured voltage  $U_A$ , indirect approach was adopted for both types of electrode configurations shown in Figure 4.9. The accuracy of the measured value is compared with the calculated value in the following sections.

### 5.5.1 Capacitive controlled rod-rod gap

#### 5.5.1.1 Lightning impulse voltage:

Figure 5.14 gives the percentage of measured lightning impulse transferred voltage  $U_A$  as a function of damping resistance obtained for various values of the supporting capacitance  $C_2$  for capacitive controlled rod-rod gap. This voltage was measured using damped capacitive voltage divider as well as resistive voltage divider for both lightning and switching impulse voltages. The magnitude of voltage  $U_A$  does not decrease with the increase in supporting capacitance  $C_2$  (for a given value of damping resistance  $R$ ) theoretically this is not correct, the voltage  $U_A$  should have decreased with the increase in supporting capacitance  $C_2$ . However, because of the measurement difficulty the voltage  $U_A$  remained the same on account of the divider capacitance and/or stray capacitance to earth.

On account of the above mentioned difficulty in the meas-

Capacitive controlled rod - rod gap ( figure - 4.9a)

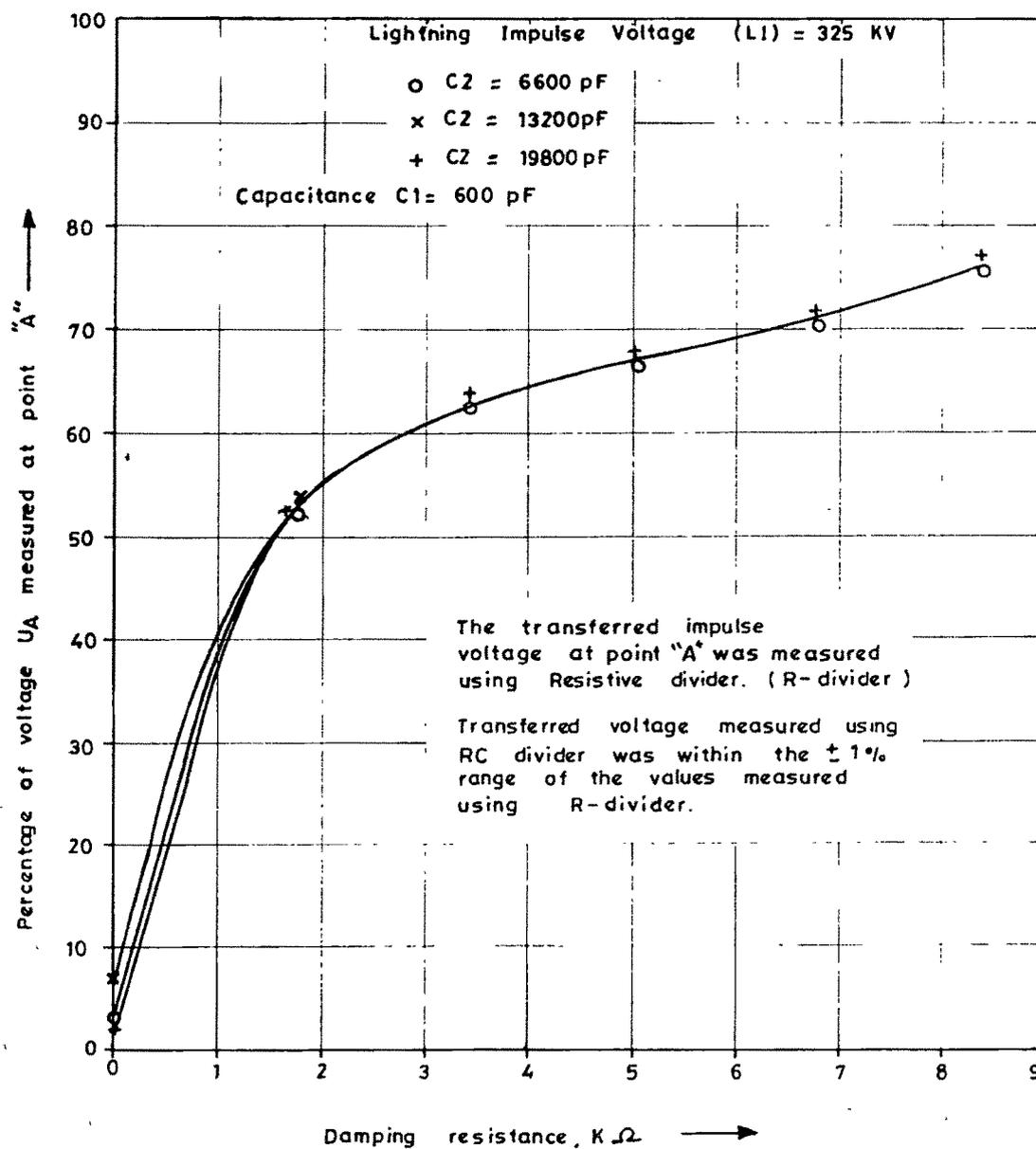


Figure. 5.14 Variation of transferred voltage  $U_A$  (LI) as a function of damping resistance,  $C_1 = 600$  pF.

urement of the transferred impulse voltage  $U_A$ , indirectly the voltage  $U_A$  was calculated based on  $U_{50}$  voltage obtained for various combinations of supporting capacitance  $C_2$  and the damping resistance. For damping resistance equal to zero Ohms, the percentage of the transferred impulse voltage at point A or B can be calculated as follows:

$$\text{Percentage of transferred impulse voltage } U_B = \frac{C_1}{C_1 + C_2 + C_{cvd}} \times 100\% \dots\dots\dots(5.12)$$

Where  $C_1$  = Test object capacitance in pF  
 $C_2$  = Supporting capacitance in pF  
 $C_{cvd}$  = Divider Capacitance in pF

For a given value of supporting capacitance  $C_2$  ( $R=0.0$  Ohms) the  $U_{50}$  value of the test object is known. It is observed that when damping resistance is connected between the test object terminal and the supporting capacitance  $C_2$ , the  $U_{50}$  increases. Theoretically this increase in  $U_{50}$  with resistance is on account of the different voltage distribution across the test object capacitance  $C_1$ , and the series combination of damping resistance and the supporting capacitance. Hence the difference in  $U_{50}$  with and without damping resistance plus the transferred impulse voltage given by equation (5.12) should be the voltage of point A, i.e. voltage  $U_A$  when damping resistance is connected between the terminals of the test object and the supporting capacitance. Figure 5.15 gives the percentage of lightning impulse transferred voltage  $U_A$  calculated based on  $U_{50}$  voltages for

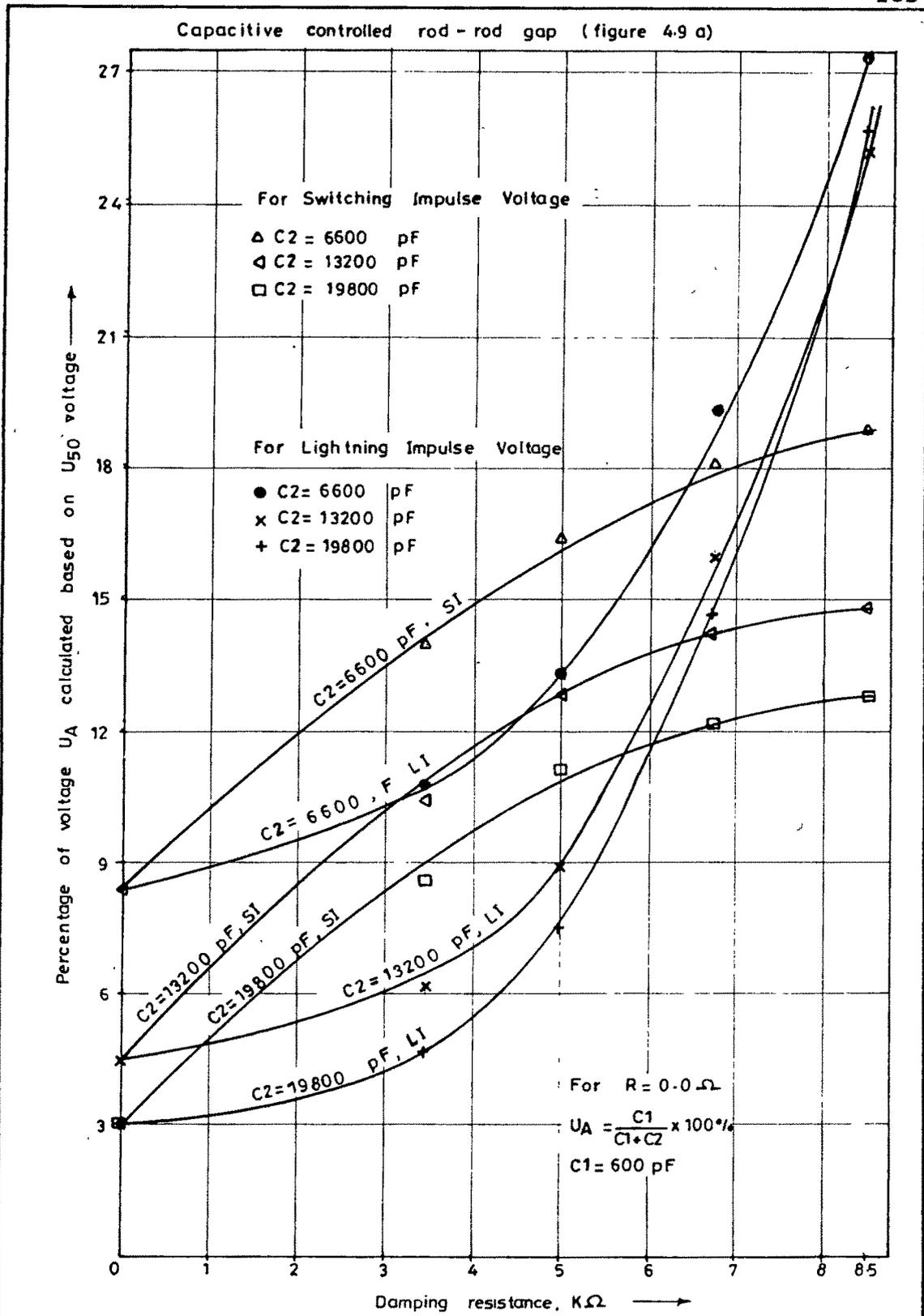


Figure 5.15 Variation of transferred voltage  $U_A$  calculated based on  $U_{50}$  as a function of damping resistance,  $C_1 = 600$  pF.

various values of damping resistance and supporting capacitance. The calculated voltage  $U_A$  increases with the increase in damping resistance and decreases with the increase in supporting capacitance. The percentage increase of lightning impulse transferred voltage at point A is of the order of 19.0 to 22.6% when damping resistance of 8.56K~~1b~~ Ohms is connected. The percentage rate of increase of voltage at point A is much high with the increase in damping resistance. It may be concluded that for lightning impulse bias testing, it is incorrect to connect the damping resistor between the terminals of the circuit breaker and the supporting capacitance  $C_2$ .

#### 5.5.1.2 Switching impulse voltage

Figure 5.16 gives the percentage of switching impulse transferred voltage  $U_A$  measured at point 'A' using damped capacitive voltage divider as a function of damping resistance  $R$  for various values of  $C_2$ . The results obtained are little better than the case of lightning impulse voltage. In figure 5.15 the calculated percentage of switching impulse transferred voltage  $U_A$  for various values of damping resistance  $R$  and capacitance  $C_2$  is also given.

It is seen from the results reported in Figure 5.15 and 5.16 that for supporting capacitance of 6600pF, the calculated and measured values of voltage  $U_A$  compare very well. However, with the increase in supporting capacitance  $C_2$  the difference in the measured and calculated values of voltage  $U_A$  increases with the increase in damping resist-

## Capacitive controlled rod - rod gap (figure. 4.9 a)

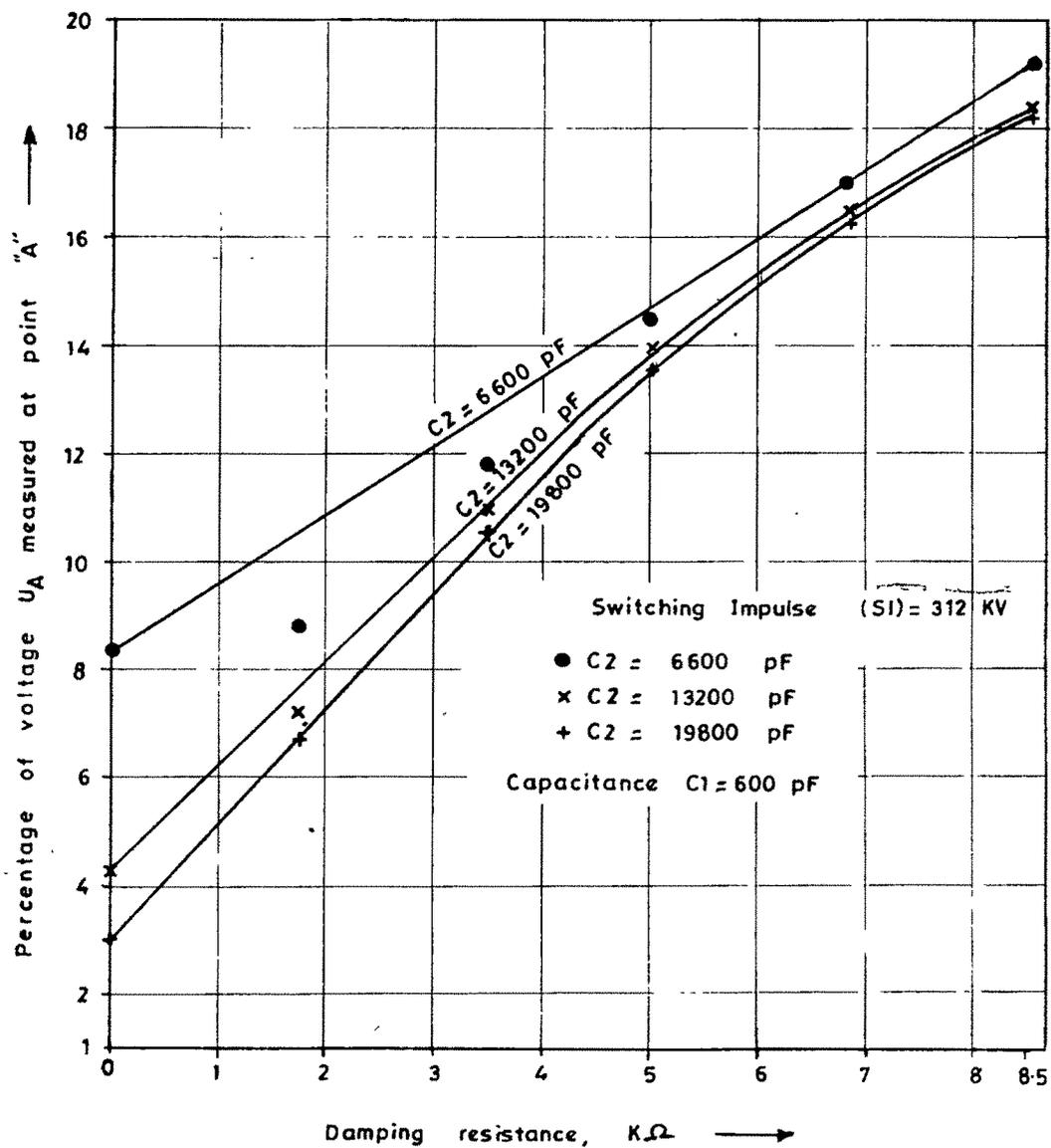


Figure. 5.16 Variation of transferred voltage  $U_A$  (SI) as a function of damping resistance,  $C_1 = 600$  pF.

ance. The measured value of switching impulse transferred voltage  $U_A$  when damping resistance of 8.56 Kilo Ohms is connected is approximately 0.7, 3.6 and 5.4% higher compared to calculated value when supporting capacitance is equal to 6600 pF, 13200 pF and 19800 pF respectively.

As discussed earlier, when divider is connected at terminal A to measure voltage  $U_A$ , the circuit configuration is changing and also while measuring lightning impulse transferred voltage  $U_A$  (refer figure 5.14) un-usual relationship between measured voltage  $U_A$  with damping resistance is observed. For measuring  $U_{50}$  values no element was connected in the circuit at terminal A. Hence, the difference between the measured and calculated voltage  $U_A$  is on account of the error in the measurement of voltage  $U_A$ . This measuring error is much lower as compared to the measuring error observed in the measurement of lightning impulse transferred voltage  $U_A$ .

### 5.5.2 Rod-rod gap

#### 5.5.2.1 Lightning impulse voltage

Figure 5.17 gives a set of curves of the percentage of lightning impulse transferred voltage  $U_A$  as a function of damping resistance for various values of supporting capacitance. With the increase in damping resistance  $R$ , the voltage  $U_A$  is increasing and it decreases with the increase in supporting capacitance. The curves for various values of supporting capacitance  $C_2$  are parallel to each other, but the percentage increase in voltage  $U_A$  is not linear with the increase in damping resistance. Figure 5.18 gives

Rod - Rod gap configuration ( figure 4.9 b)

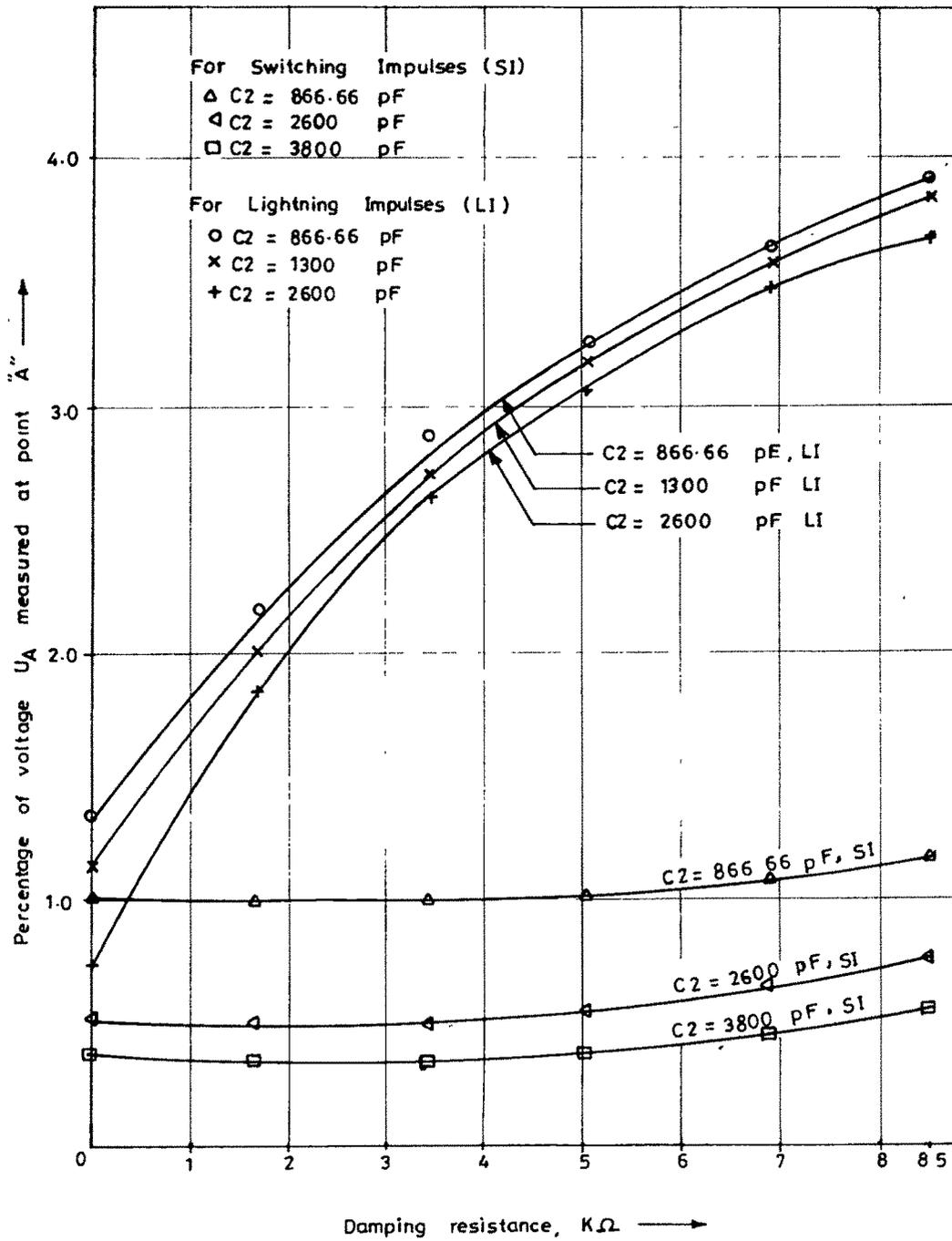


Figure. 5.17 Variation of transferred voltage  $U_A$  (LI and SI) as a function of damping resistance, rod-rod gap.

Rod - Rod gap Configuration (figure 4.9b)

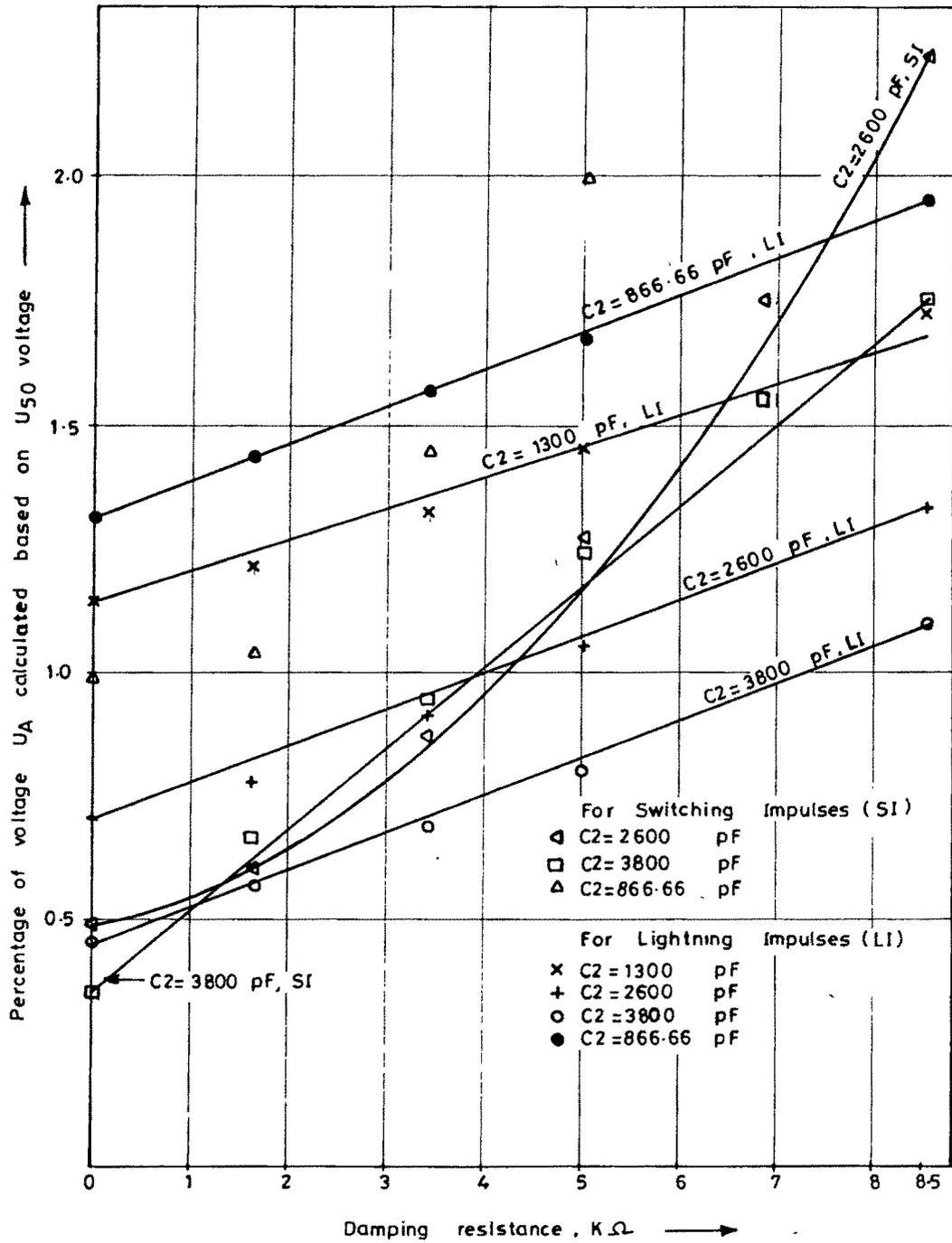


Figure. 5.18 Variation of transferred voltage  $U_A$  calculated based on  $U_{50}$  as a function of damping resistance, rod - rod gap.

a set of curves of percentage of lightning impulse transferred voltage  $U_A$  calculated based on the  $U_{50}$  values as a function of damping resistance  $R$  for various values of supporting capacitance  $C_2$ . The calculated value of voltage  $U_A$  increases linearly with the increase in damping resistance. The measured voltage at point A when a damping resistance of 8.56 Kilo Ohms is connected is 3.8 per cent of the applied impulse voltage ( $C_2 = 866.66\text{pF}$ ) while the calculated value of voltage  $U_A$  is 1.97 per cent, i.e. nearly fifty per cent of the measured value.

#### 5.5.2.2 Switching impulse voltage

Figure 5.17 also gives a set of curves of the percentage of switching impulse transferred voltage  $U_A$  as a function of damping resistance for various values of supporting capacitance. In this case the rate of increase of voltage  $U_A$  with the increase in damping resistance is very small as compared to the case of lightning impulse transferred voltage  $U_A$ . The calculated switching impulse transferred voltage  $U_A$  obtained based on  $U_{50}$  values for various combinations of damping resistance and supporting capacitance is given in Figure 5.18. The measured voltage  $U_A$  when a damping resistance of 8.56 Kilo Ohms is connected is approximately 0.8% ( $C_2 = 2600\text{pF}$ ), while the calculated voltage  $U_A$  is 2.23%.

#### 5.6 Reduction of transferred impulse voltage $U_B$

As discussed in Chapter-3, because of the capacitive coupling between the test object capacitance and the capacitance connected on the AC terminal, part of the applied

impulse voltage to terminal D of figure 4.1, gets transferred on the AC terminal. This causes voltage distortion on the AC wave. Because of the opposite polarity of the impulse voltage, the magnitude of the AC voltage applied at the instant of impulse voltage reduces. On account of this, the test object gap is not stressed to the desired test voltages. The voltage depression on AC wave is equal to the transferred impulse voltage on AC terminal. Hence, to stress the test object gap to the required test voltages, the magnitude of power frequency voltage is increased by an amount equal to the magnitude of the transferred impulse voltage on AC terminal.

If the circuit connections for bias voltage test is made as shown in Figure 4.1, one will measure the voltage  $U_B$  and compensate for the voltage drop on AC terminal. If this is the case, it would be necessary to study the reduction of the transferred impulse voltage  $U_B$  caused by incorrect location of the damping resistor. Since there is no problem in the measurement of voltage  $U_B$  by using damped capacitive voltage divider, the percentage reduction in voltage  $U_B$  is derived from the results reported in Chapter-4.

Figure 5.19 gives a set of results of percentage decrease in voltage  $U_B$  as a function of damping resistance for various values of supporting capacitance. The test object capacitance was 600pF. As seen from this figure, in the case of switching impulse voltages, the voltage  $U_B$  does not vary with the insertion of damping resistance, it

Capacitive controlled rod-rod gap ( figure 4.9a)

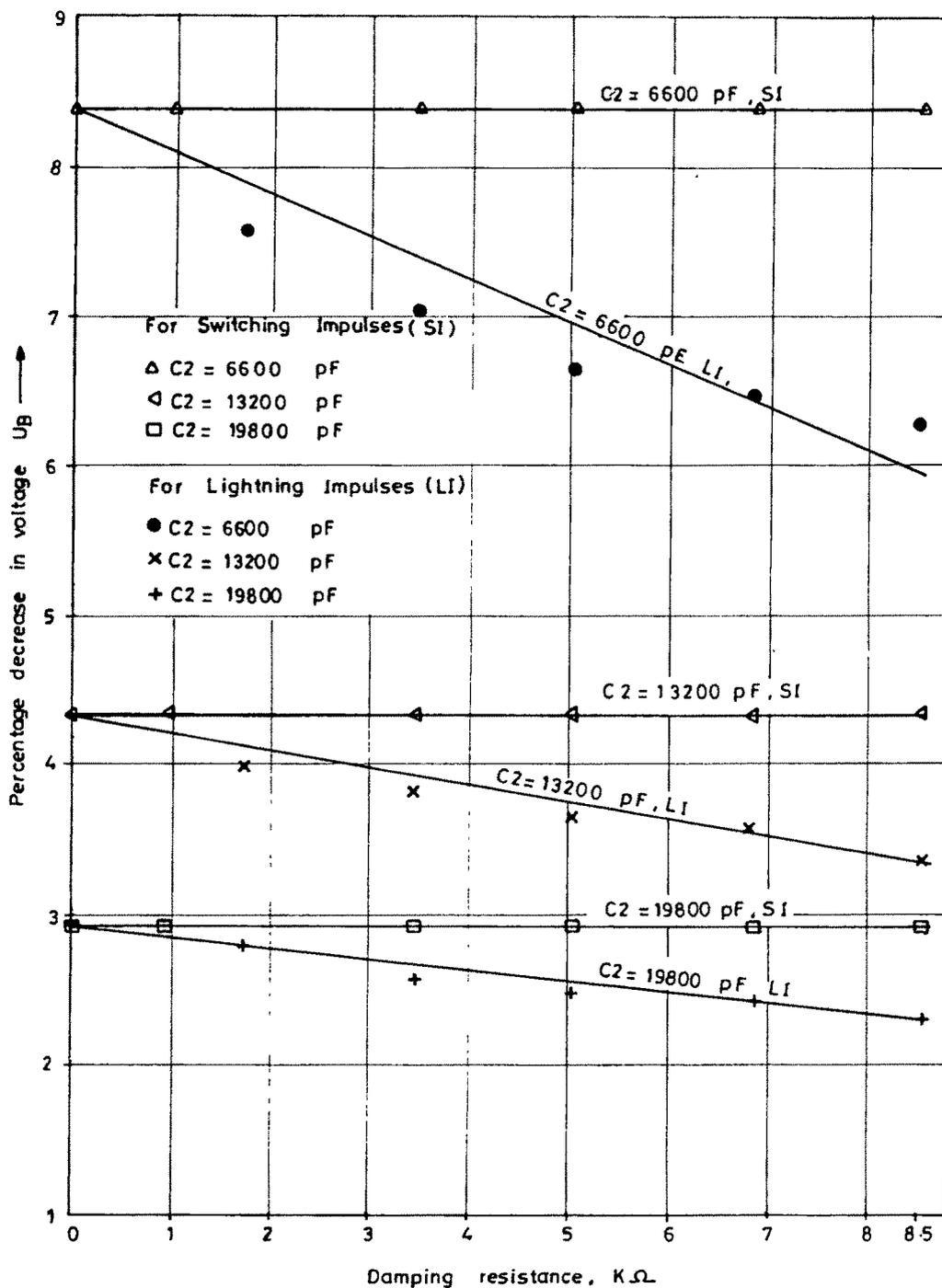


Figure. 5.19 Variation of transferred voltage  $U_B$  as a function of damping resistance,  $C_1 = 600$  pF.

gives a straight line relationship. While in the case of lightning impulse voltage the voltage  $U_B$  decreases with the increase in damping resistance. The percentage decrease in lightning impulse transferred voltage  $U_B$  is approximately 2.39, 1.0 and 0.6 for supporting capacitance  $C_2$  equal to 6600pF, 13200pF and 19800pF respectively when damping resistance of 8.56 Kilo Ohms is connected between the test object terminal and the supporting capacitor  $C_2$ .

Figure 5.20 gives a set of results of percentage decrease in voltage  $U_B$  as a function of damping resistance for rod-rod gap electrode configuration. For this type of electrode configuration also, voltage  $U_B$  does not vary for the case of switching impulse voltages by increasing the value of damping resistance. While in the case of lightning impulse voltages, the voltage  $U_B$  decreases with the increase in damping resistance. The percentage decrease in voltage  $U_B$  is approximately 0.21, 0.265 and 0.19 for supporting capacitance  $C_2$  equal to 866.66pF, 1300pF and 3800pF respectively when damping resistance of 8.56Kilo Ohms is connected between the test object terminal and the supporting capacitor  $C_2$ .

The percentage reduction of the transferred impulse voltage  $U_B$  because of damping resistance in the case of disconnector type open gap is almost negligible. However, in the case of capacitive controlled rod-rod gap, for lightning impulse bias test; the error introduced in the measurement of voltage  $U_B$  is large and cannot be neglected. For the test object capacitance equal to 600pF, the required value of the supporting capacitance will be of the order of 8800pF, for this value of

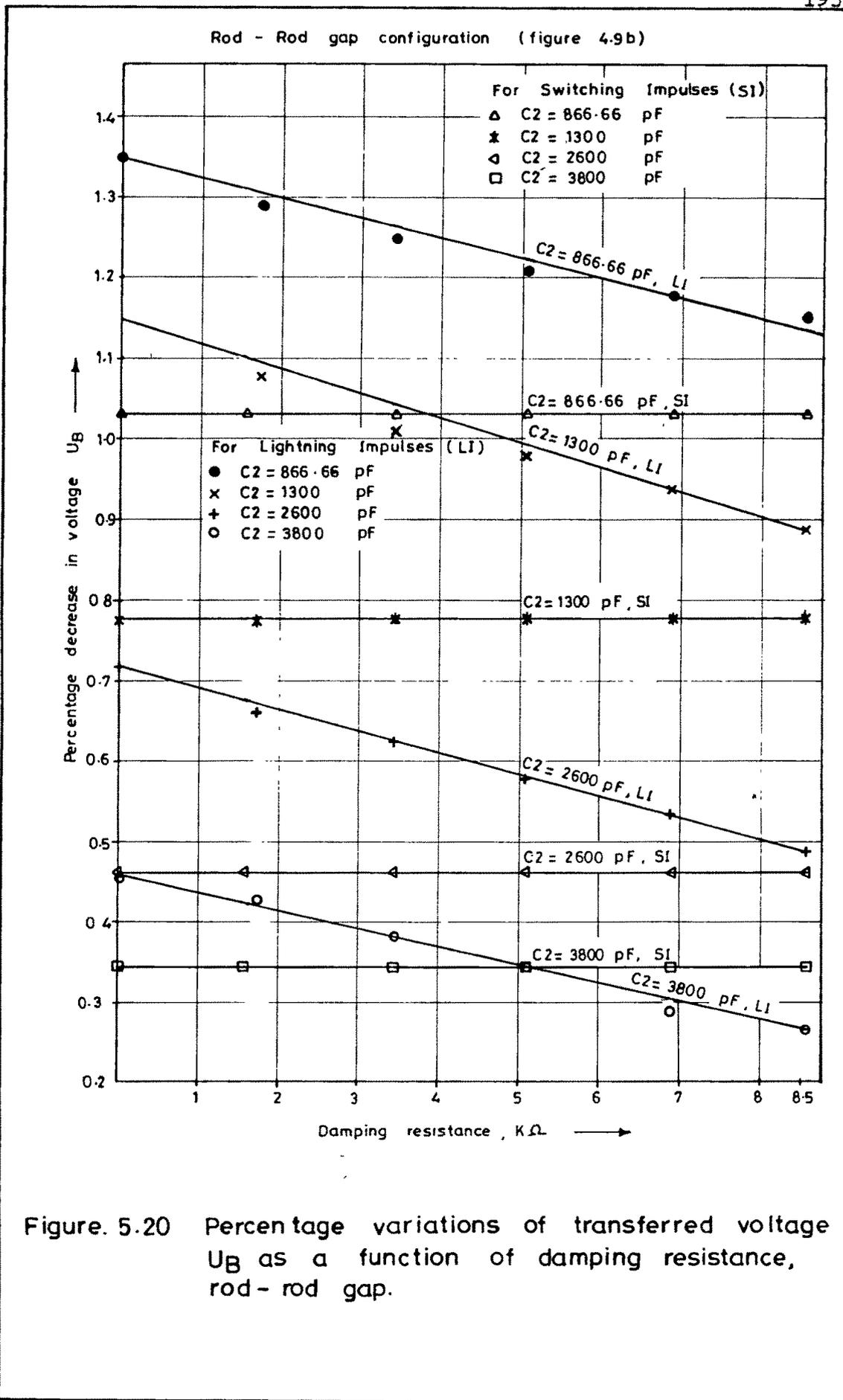


Figure 5.20 Percentage variations of transferred voltage  $U_B$  as a function of damping resistance, rod-rod gap.

supporting capacitor  $C_2$ , the error in the measurement of voltage  $U_B$  will be of the order of 1.44% when damping resistance of 8.56 Kilo Ohms is connected between the terminals of the test object and the supporting capacitance.

#### 5.7 Total reduction of voltage stress across the test object

It has been established now and proven that the correct test circuit for conducting bias test should be as shown in Figure 4.2. However, if the bias test circuit shown in Figure 4.1 is used, one should know the total error introduced in applying the test voltages to the test object. The error introduced is on account of the following two main aspects:

- a. The voltage distribution is affected by connecting a damping resistor between the test object terminal and the supporting capacitance. Higher the value of damping resistance, lesser is the voltage applied across the test object (error  $e_1$ ).
- b. The error introduced in compensating for the transferred impulse voltage on AC terminal i.e. the error in the measurement of voltage  $U_B$ , which is ultimately added to the required power frequency test voltage. This error also increases with the increase in the value of damping resistor (error  $e_2$ ).

For lightning and switching impulse voltages the total

error can be easily calculated by using the results given in Figures 5.5 and 5.19 for capacitive controlled rod-rod gap. In the case of lightning impulse voltage, the error  $e_1$  for  $C_2 = 6600\text{pF}$  and  $R = 8.56 \text{ Kilo Ohms}$  is of the order of 19 per cent of the applied lightning impulse voltage and the error  $e_2$  is 2.39% of the power frequency voltage.

$$\begin{aligned} \text{Total error in kV} &= e_1 + e_2 \\ &= \text{impulse voltage} \times \frac{19}{100} \\ &+ \frac{\text{power frequency}}{\text{peak voltage}} \times \frac{2.39}{100} \end{aligned}$$

The error  $e_1$  mentioned above is derived based on the  $U_{50}$  values of the test object with and without damping resistance connected in the circuit. The error  $e_2$  is obtained by measuring the voltage  $U_B$  with and without damping resistance. When the withstand voltage test is conducted on the longitudinal insulation, the gap will be stressed less by an amount of voltage equal to the total error  $e_1 + e_2$ .

Thus,

$$\begin{aligned} \text{Actual Voltage Stress across the breaker is} &= \text{Impulse voltage} + \text{power frequency peak voltage} - \text{Total error in kV} \end{aligned}$$

The corresponding errors introduced in the case of switching impulse voltage is of the order of 10 per cent (error  $e_1$ ). The error in the measurement of voltage  $U_B$  is almost negligible (i.e. error  $e_2 = 0.0\%$ ).

Similarly, the total error for the case of rod-rod gap can be derived by using the results given in Figure 5.9 and 5.20. For lightning impulse voltages the error  $e_1$  for

$C_2 = 866.66\text{pF}$  and  $R = 8.56$  Kilo Ohms is of the order of 0.65 per cent of the lightning impulse voltage and error  $e_2$  is 0.21 per cent of the power frequency test voltage.

The corresponding errors introduced in the case of switching impulse voltage is of the order of 2.3 per cent (error  $e_1$ ). The error in the measurement of voltage  $U_B$  is almost negligible (i.e. error  $e_2 = 0.0$ ).

### 5.8 Standard deviation sigma and their variances

The estimation for calculation of standard deviation sigma was proposed by Dixon and Mood/150/. The formulae for calculating sigma, variance of  $U_{50}$  and variance of sigma are given in section 5.2. The measured standard deviation  $\hat{\sigma}$  will converge to the true value  $\sigma_0$  when the number of voltage applications is infinite. However, the measured standard deviation is biased when the number of voltage applications are finite. As described earlier the number of impulses applied to the rod-rod gap and capacitive controlled gap for obtaining  $U_{50}$  were fifty for both lightning and switching impulses. The value of standard deviation  $\hat{\sigma}$ , the variance of  $U_{50}$  and the variance of  $\hat{\sigma}$  are summarised in Table 5.1 to 5.3 for various values of damping resistance and supporting capacitance  $C_2$ . The voltage step for lightning impulses was taken as 3 per cent of the approximately estimated  $U_{50}$  value and for switching impulses the voltage step was taken as 6 per cent of the approximately estimated  $U_{50}$  value. However, inspite of accurate estimation of starting voltage  $U_0$  for many cases it was observed

TABLE: 5.1

Standard deviation  $\hat{\sigma}$ , the variance of  $U_{50}$  and the variance of  $\hat{\sigma}$  for capacitive controlled rod-rod gap for switching impulse voltages,  $C_1=600\text{pF}$ .

Sr. No.	Capacitance $C_2$ $\mu\text{F}$	Damping resistance Kilo-Ohms	$U_{50}$ kV	$\frac{\hat{\sigma}}{U_{50}}\%$	$\sqrt{\frac{\text{Var}50}{U_{50}}}\%$	$\sqrt{\frac{\text{Var}\hat{\sigma}}{U_{50}}}\%$
1	6600	0.0	640.4	3.285	0.7698	0.918
2	6600	5.05	690.9	2.8658	0.6568	0.8074
3	6600	6.84	702.5	2.055	0.4893	0.5698
4	6600	8.56	707.2	3.0288	0.6971	0.8465
5	13200	0.0	629.79	2.4198	0.5732	0.6709
6	13200	5.05	680.9	1.965	0.4714	0.5469
7	13200	6.84	690.4	2.5738	0.5984	0.7163
8	13200	8.56	693.09	2.155	0.5733	0.5973
9	19800	0.0	620.0	3.556	0.84	1.0196
10	19800	5.05	671.5	5.003	1.1075	1.5215
11	19800	6.84	675.35	3.839	0.8714	1.0898
12	19800	8.56	681.0	3.826	0.8727	1.0909

TABLE:5.2

Standard deviation  $\delta$ , the variance of  $U_{50}$  and the variance of  $\delta$  for capacitive controlled rod-rod gap for lightning impulse voltages,  $C_1=600\text{pF}$ .

Sr. No.	Capacitance $C_2$ - $\mu\text{F}$	Damping resistance Kilo Ohms	$U_{50}$ kV	$\frac{\delta}{U_{50}}$ %	$\frac{\sqrt{\text{Var}50}}{U_{50}}$ %	$\frac{\sqrt{\text{Var}\delta}}{U_{50}}$ %
1	--	0.0	564.34	1.8527	0.4001	0.4607
2	6600	0.0	644.6	2.3487	0.4837	0.5885
3	6600	5.05	672.5	2.7296	0.554	0.7003
4	6600	6.84	718.5	3.7649	0.7398	1.114
5	6600	8.56	768.17	6.3	1.1833	2.369
6	13200	0.0	626.2	1.805	0.3848	0.4471
7	13200	5.05	652.2	2.3377	0.48923	0.5887
8	13200	6.84	726.5	--	--	--
9	13200	8.56	754.0	3.022	0.6034	0.7997
10	19800	0.0	608.5	1.857	0.39917	0.4601
11	19800	5.05	629.3	2.297	0.4709	0.5812
12	19800	6.84	683.0	1.8169	0.3931	0.4598
13	19800	8.56	749.0	3.4325	0.6658	1.016

**TABLE 5.3**

Standard deviation  $\hat{\sigma}$ , the variance of  $U_{50}$   
 and the variance of  $\hat{\sigma}$ , for rod-rod gap  
 electrode configuration for switching  
 impulse waves

Sr. No.	Capacitance $C_2$ $\mu$ F	Damping Resistance Kilo Ohms	$U_{50}$ KV	$\frac{\hat{\sigma}}{U_{50}}$ %	$\sqrt{\frac{\text{Var } 50}{U_{50}}}$ %	$\sqrt{\frac{\text{Var } \hat{\sigma}}{U_{50}}}$ %
1	866.66	0.0	327.47	3.148	0.6675	0.783
2	866.66	1.716	327.54	3.553	0.74189	0.8811
3	866.66	3.64	328.75	3.5923	0.7476	0.8922
4	866.66	5.05	332.76	3.9968	0.8192	1.0067
5	866.66	8.56	334.0	3.1209	0.6491	0.7784
6	0.0	0.0	304.0	2.9296	0.638	0.7322
7	2600	0.0	327.19	3.334	0.6901	0.8336
8	2600	1.716	328.09	3.374	0.7169	0.83513
9	2600	3.46	328.77	2.8697	0.6195	0.7172
10	2600	5.05	329.5	2.4613	0.5462	0.6191
11	2600	8.56	330.347	2.873	0.609	0.7122
12	3800	0.0	324.7	4.33	0.87125	1.127
13	3800	1.716	326.5	3.231	0.685	0.7993
14	3800	5.05	328.5	3.247	0.681	0.8066
15	3800	8.56	330.0	3.1587	0.656	0.7833

that the ratio  $\frac{d}{\hat{\sigma}}$  was either less than 0.3 or greater than 1.7. For such cases, the voltage step  $d$  was changed and the test was repeated to derive  $U_{50}$  and  $\hat{\sigma}$  to satisfy the condition  $0.3 < \frac{d}{\hat{\sigma}} < 1.7$ . Whenever, the test was repeated, the value  $U_{50}$  remained more or less same, only the number of voltage steps applied were changing based on the chosen voltage step- $d$ . The average value of sigma in case of switching impulse waves were 2.808, 2.276 and 4.056 for capacitance  $C_2$  equal to 6600, 13200 and 19800 pF respectively for test object capacitance equal to 600pF. For lightning impulse voltages the average value of sigma were 3.789, 2.38 and 2.35 for supporting capacitance  $C_2$  equal to 6600, 13200 and 19800pF respectively for test object capacitance equal to 600pF. For rod-rod gap, the average value of sigma in case of switching impulse voltages were 3.4822, 2.9736 and 3.4916 for supporting capacitance equal to 866.66, 2600 and 3800pF respectively.

#### 5.9 Protection of high voltage testing transformer

The protection of high voltage testing transformer in case flashover occurs across the longitudinal insulation was discussed at depth in Chapter-3. It was concluded that the time constant of the circuit should be increased in order to reduce the severity of the chopped voltage by connecting high voltage damping resistor between the terminals of the test object and the supporting capacitance  $C_2$ .

Higher the value of damping resistor, the severity of the chop reduces. However, the conclusion drawn in Chapter-4 is contradictory to what has been discussed in Chapter-3.

From the testing point of view and as per the systems requirement, the damping resistance should be connected between the terminals of the high voltage testing transformer and the supporting capacitance. Then the question arises, how to protect the high voltage testing transformer? By connecting the damping resistance between the high voltage test transformer and the supporting capacitance, still the time constant of the circuit can be increased by increasing the value of  $R$ . However, one cannot go on increasing the value of damping resistance because of the following reasons.

1. The  $I^2R$  losses increases with increase in  $R$
2. The sphere gap is used to protect the High Voltage testing transformer from high transient recovery over-voltages. When the open gap flashes over the sphere gap will flashover, and because of high short circuit current, the relay will operate and power supply to the testing transformer will be disconnected.

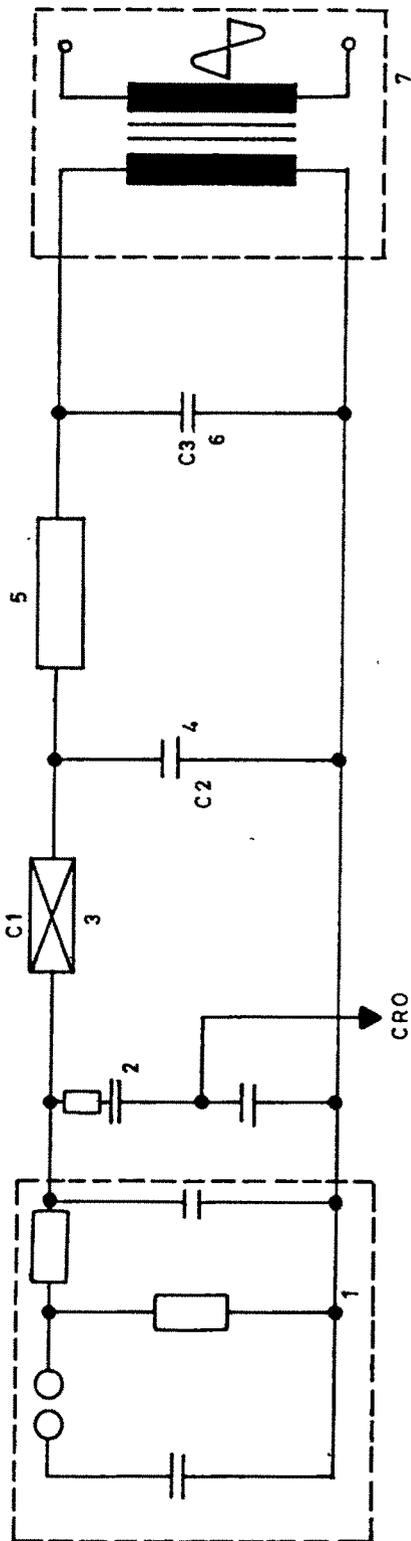
However, if the value of damping resistance  $R$  is very high, the short circuit current may not be sufficient for the relay to operate, under such circumstances, the transformer will remain under dead short circuit for longer time till the operator trips it manually. Also, because of high impedance, premature arc extinction may take place because of insufficient ionisation of the arc path which causes overvoltages upto twice the test voltage value. This is very dangerous

and should be avoided.

3. The dimensions of the high voltage damping resistance will increase with the increase in resistance, the laboratory dimensions may not permit this.

Because of the above mentioned difficulties it was decided to increase the time constant of the circuit by connecting high voltage capacitor  $C_3$  as shown in Figure 5.21. The capacitor  $C_3$  can be connected only if the high voltage testing transformer has sufficient current rating.

Thus, by connecting an additional capacitor  $C_3$ , the protection problem of winding insulation of the high voltage testing transformer can be sorted out, but one has to study the effect of capacitance  $C_3$  on the voltage distribution across the longitudinal insulation. It was aimed to study this by finding  $U_{50}$  with and without capacitor  $C_3$ . Table 5.4 gives  $U_{50}$  values for rod-rod gap and capacitive controlled rod-rod gap electrodes configuration. The maximum difference in  $U_{50}$  with and without capacitor  $C_3$  is 0.288 per cent. It may be concluded that the capacitor  $C_3$  has no effect on  $U_{50}$  of the test object and hence there is no harm in connecting capacitor  $C_3$  for the protection of high voltage testing transformer. The value of  $C_3$  will be decided based on the current capacity of the high voltage testing transformer. Higher the value of  $C_3$ , lesser is the severity of the chopped wave.



1. Impulse Voltage generator 2200 KV, 66 kJ
2. Damped capacitive voltage divider 2000 KV, 713 pF
3. Test objet (either Rod - Rod gap or Capacitive controlled gap. see figure 4.9 )
4. Supporting capacitor C2
5. Damping Resistor 8.56 K $\Omega$
6. Capacitance for the protection of H.V. Testing Transformer (C3)
7. Power frequency Transformer 750 KV, 750 KVA

Figure. 5.21 Circuit layout to study the effect of capacitance C3 on the flashover voltage  $U_{50}$  of the test object.

TABLE:5.4

U<sub>50</sub> Voltage with and without capacitor C<sub>3</sub>

Sr. No.	Electrode configur- ation	Impulse wave	U <sub>50</sub> ,kV		% variations $\frac{U_{50A}-U_{50B}}{U_{50A}} \times 100$ %
			with C <sub>3</sub> U <sub>50A</sub>	without C <sub>3</sub> U <sub>50B</sub>	
1	Rod-rod gap	Lightning impulse C <sub>2</sub> =1200pF C <sub>3</sub> =1300pF	458.72	457.70	0.222
2	Rod-rod gap	Switching impulse C <sub>2</sub> =1200pF C <sub>3</sub> =1300pF	330.82	331.62	- 0.242
3	Capaciti- ve cont- rolled gap	Lightning impulse C <sub>2</sub> =13200pF C <sub>3</sub> =13200pF	625.0	626.30	- 0.208
4	Capaci- tive contro- lled gap	Switching impulse C <sub>2</sub> =6600pF C <sub>3</sub> =6600pF	642.15	644.00	- 0.288

### 5.10 Conclusion

The critical flashover voltage  $U_{50}$  was derived for capacitive controlled rod-rod gap and rod-rod gap electrode configuration for both lightning and switching impulse waves. The value of damping resistor  $R$  and capacitance  $C_2$  was varied, and few sets of results were obtained and analysed. The procedure for establishing the correct value of lightning impulse and switching impulse transferred voltage  $U_A$  and the total reduction of voltage stress across the longitudinal insulation when damping resistance is connected between the terminals of the test object and supporting capacitance is discussed.

Some of the important conclusions drawn from this study are as follows:

1.  $U_{50}$  lightning impulse voltage increases with increase in damping resistance and it decreases with the increase in supporting capacitance  $C_2$  for capacitive controlled rod-rod gap. The percentage increase of  $U_{50}$  is as high as 23 per cent when damping resistor of 8.56 Kilo Ohms is connected between the terminals of the test object and the supporting capacitance. The increase in  $U_{50}$  value is not linear with the increase in damping resistance.
2.  $U_{50}$  switching impulse voltage also increases with increase in damping resistance and it decreases with increase in supporting capacitance for capacitive controlled rod-rod gap. The percentage increase in  $U_{50}$  for three values of supporting capacitance

6600pF, 13200 pF and 19800pF is fairly the same for a given value of damping resistance. The increase in  $U_{50}$  voltage is approximately 10 per cent when damping resistor of 8.56 Kilo Ohms is connected between the terminals of the test object and the supporting capacitance  $C_2$ .

3. It seems when damping resistance  $R$  is greater than 7 Kilo Ohms, the increase in  $U_{50}$  switching impulse voltage is reaching to a saturation level, i.e. further increase in damping resistance does not lead to increase in  $U_{50}$  for capacitive controlled rod-rod gap.
4.  $U_{50}$  lightning impulse voltage increases linearly with increase in damping resistance for rod-rod gap. The percentage increase in  $U_{50}$  is maximum 0.7 per cent when damping resistance of 8.56 Kilo Ohms is connected between the terminals of the test object and the supporting capacitance  $C_2$ . This is negligible as compared to the capacitive controlled rod-rod gap.
5. While conducting bias voltage test on rod-rod gap the damping resistor may be connected between the terminals of the test object and supporting capacitance, so that the severity of the chopped wave is not dangerous for the winding of the High Voltage test transformer.

6.  $U_{50}$  switching impulse voltage also increases with increase in damping resistance for rod-rod gap. The percentage increase in  $U_{50}$  is between 1.3 and 2.4 per cent when damping resistance of 8.56 Kilo Ohms is connected between the test object terminal and the supporting capacitance  $C_2$  ranging from 866.66pF to 3800pF.
7. The measured values of  $U_{50}$  for capacitive controlled rod-rod gap are approximately 4.5 per cent more as compared to the calculated  $U_{50}$  for both lightning and switching impulse voltages. This is attributed to the different field configurations because of change in circuit connections.
8. The measured values of  $U_{50}$  for rod-rod gap are approximately 3.8 per cent more as compared to calculated  $U_{50}$  for lightning impulses and 6.5 per cent for switching impulses.
9. For capacitive controlled rod-rod gap, it is not possible to measure the lightning impulse transferred voltage  $U_A$  accurately even by using damped capacitive voltage divider.
10. The percentage increase of lightning impulse transferred voltage  $U_A$  is of the order of 23 per cent when damping resistance of 8.56 Kilo ohms is connected between the terminals of the test object and supporting capacitor for capacitive controlled rod-rod gap. Hence, it is incorrect to connect a damping resistor

between the circuit breaker terminal and the supporting capacitor while conducting bias test.

11. For capacitive controlled rod-rod gap the calculated and measured values of switching impulse transferred voltage  $U_A$  differs only between 0.7 and 5.4% for a supporting capacitance  $C_2$  ranging from 6600 pF to 19800 pF when damping resistance is varied from 0.0 Ohms to 8.56 Kilo Ohms. This variations may be attributed to the measurement problem.
12. The measured lightning impulse transferred voltage  $U_A$  for rod-rod gap, when a damping resistance of 8.56 Kilo Ohms is connected is 3.8 per cent of the applied impulse voltage ( $C_2 = 866.66$  pF), while the calculated voltage  $U_A$  is 1.97 per cent, i.e. nearly fifty per cent of the measured value.
13. The measured switching impulse transferred voltage  $U_A$  for rod-rod gap is more or less independent of damping resistance. However, the calculated voltage  $U_A$  increases with the increase in damping resistance.
14. If the damping resistor is connected between the terminals of the supporting capacitor and the test object, it leads to an error in the measurement of voltage  $U_B$  which is required for compensating the voltage drop on AC terminal while conducting bias voltage test. For capacitive controlled rod-rod gap when test object capacitance  $C_1$  is equal to

600pF, the error in the compensation of voltage drop will be approximately 1.44 per cent which is too large ( $C_2 = 8800\text{pF}$ ).

15. While conducting bias voltage test on circuit breakers ( $C_1 = 600\text{pF}$  and  $C_2 = 13200\text{pF}$ ) when a damping resistor of 8.56 Kilo Ohms is connected between the terminals of the test object and the supporting capacitor, the longitudinal insulation of the circuit breaker will be stressed less by approximately 20.4 per cent of the lightning impulse voltage and 1.0 per cent of the power frequency voltage while conducting lightning impulse bias test, and will be stressed less by approximately 10 per cent of the switching impulse voltage while conducting switching impulse bias test.
16. While conducting bias voltage test on rod-rod gap when a damping resistor of 8.56 Kilo Ohms is connected between the terminals of the test object and the supporting capacitor  $C_2$  ( $C_2 = 866.6\text{pF}$ ) the longitudinal insulation will be stressed less by approximately 0.65 per cent of the lightning impulse voltage and 0.21 per cent of the power frequency voltage while conducting lightning impulse bias test, and will be stressed less by approximately 2.3 per cent of the switching impulse voltage while conducting switching impulse bias test.
17. The average values of standard deviation sigma for switching impulse waves were 2.8, 2.276 and 4.056

for capacitance  $C_2$  equal to 6600, 13200 and 19800pF respectively for test object capacitance equal to 600pF, and for lightning impulse voltage, the average values of sigma were 3.39, 2.38 and 2.35 per cent for supporting capacitance  $C_2$  equal to 6600, 13200 and 19800pF respectively.

18. For rod-rod gap the average value of sigma for switching impulse voltages were 3.48, 2.97 and 3.44 per cent for supporting capacitance  $C_2$  equal to 866.66, 2600 and 3800pF respectively.
  
19. For the protection of high voltage testing transformer, an additional capacitor  $C_3$  can be connected in parallel to the High Voltage testing transformer. This will reduce the severity of the chopped wave on the testing transformer.